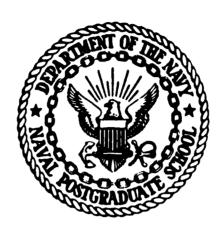


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NAVAL POSTGRADUATE SCHOOL Monterey, California



THESIS



IMAGE TEXTURE GENERATION USING AUTOREGRESSIVE INTEGRATED MOVING AVERAGE (ARIMA) MODELS

by

Steven Clifford Rathmanner

March 1987

Thesis Advisor:

Charles W. Therrien

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Image Texture Generation Using Autoregressive Integrated Moving Average (ARIMA) Models

by

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Submitted in partial fulfillment of the requirements for the degree of

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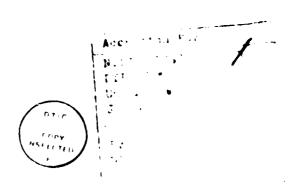
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ABSTRACT

This thesis involves investigation of linear filtering models as a means of generating texture in images. Various autoregressive filter models are used to generate various textures, and the results are analyzed to determine relationships between filter parameters and texture characteristics. A two-dimensional counterpart to the autoregressive integrated moving average (ARIMA) model from one-dimensional time series analysis theory is developed and tested for texture modeling applications. All these models are driven by white noise, and to the extent that real images can be reproduced this way, advantages in image texture transmission could be realized. Results of this work indicate that the purely autoregressive models work well for some types of image textures, but that for the textures studied the ARIMA model is not particularly suitable.



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IBM System/370 3033

TABLE OF CONTENTS

I.	INT	INTRODUCTION			
II.	THE	AUTOREGRESSIVE IMAGE MODEL 1	L 1		
	A.	ARBITRARILY SELECTED FILTER COEFFICIENTS; P = 2, Q = 21	L2		
	В.	TWO POLE, SEPARABLE AUTOREGRESSIVE MODEL	١7		
	c.	FOUR POLE, SEPARABLE AUTOREGRESSIVE MODEL 2	27		
		1. Complex Conjugate Poles 2	29		
		2. Two Real Poles	3 5		
	D.	IMAGE TEXTURE ROTATION TRANSFORMATION	10		
	E.	SUMMARY	1 1		
III.		LEMENTATION OF AN FIR SUMMATION FILTER TWO DIMENSIONS	13		
IV.		LICATION OF THE ARIMA MODEL TO IMAGE TURES	56		
	A.	APPLICATION OF LAPLACIAN INVERSE FILTER TO AUTOREGRESSIVELY GENERATED IMAGES	56		
	В.	AUTOREGRESSIVE FILTER PARAMETER ESTIMATION PROCEDURES	5 C		
	c.	APPLICATION TO REAL IMAGE TEXTURES	5]		
	D.	SUMMARY	7 4		
v.	CON	CLUSIONS 7	7 5		
APPENI	OIX A	A: COMPUTER PROGRAMS, SUBROUTINES, AND FUNCTIONS	77		
APPENI	OIX I	B: DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE TWO	٦.		

APPENDIX C:	GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE TWO POLE AUTOREGRESSIVE MODEL	106
APPENDIX D:	DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL	109
APPENDIX E:	GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL	115
APPENDIX F:	DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL (WITH TWO POLES ON THE REAL AXIS)	120
APPENDIX G:	GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL (WITH TWO POLES ON THE REAL AXIS)	124
APPENDIX H:	LAPLACIAN INVERSE FILTER FORMS	127
APPENDIX I:	CONVOLUTION OF LAPLACIAN DIFFERENCE OPERATOR AND VARIOUS SIZE FIR INVERSE FILTERS	129
LIST OF REFE	RENCES	130
TNTTTAL DIST	RIRITION LIST	121

I. <u>INTRODUCTION</u>

The purpose of this thesis is to investigate the types and quantity of image textures generated using a twodimensional (2-D) extension of the Autoregressive Integrated Moving Average (ARIMA) model. For the one-dimensional (e.g., time series) case, the theories and formulas describing this model are outlined in Box and Jenkins [Ref. 1:pp. 85-103]. The 1-D ARIMA model is useful when the time series modeled is not stationary but exhibits some homogeneity in the sense that, except for statistical differences between parts of the time series, different parts of the process behave similarly. cases some suitable difference of the process may stationary, and hence may be accurately modeled by Autoregressive Moving Average (ARMA) or purely Autoregressive (AR) model. The resulting stationary time series (generated by an appropriate ARMA or AR filter with white noise input) is applied to an integra-tion or summation filter (the inverse of the difference operation) to generate the original nonstationary time series. 1:p. 85] Figure 1-1 shows a block diagram of this process.

This work attempts to extend these concepts to two-dimensional signal processing. In order to simplify the model, the moving average (MA) portion of it will be

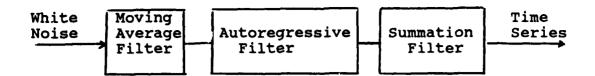


Figure 1-1 Block Diagram for the Autoregressive Integrated Moving Average Model

eliminated (i.e., no zeros in the filter Z transform) so that only purely AR models will be considered for stationary image generation. The procedures for modeling image textures using AR models with white noise input are well established [Ref. 2:pp. 454-456]. However, a suitable two-dimensional difference operation and its inverse must be found to implement the concepts outlined above.

The research is divided into four areas:

- 1) Investigation of the various types of image textures generated using AR models where filter coefficients and size are determined a) arbitrarily, b) using a two-pole separable model, and c) using a four-pole separable model. Separability refers to the fact that the Z transform of the AR filter can be factored into components representing each dimension or direction of the image.
- 2) Selection of a difference operator and a realizable inverse (integration or summation) filter.
- 3) Application of the above autoregressively generated images to the summation filter, and evaluation of these results.
- Attempted reproduction of actual images textures using AR models whose coefficients are determined using the statistics of the image, and comparison of these results to those obtained by a) applying the difference operator to the real image, b) finding the coefficients of the AR model that reproduce the difference image, and c) applying the difference image to the summation filter. This comparison was intended

to discover what improvements, if any, may be realized using the ARIMA model vice a purely ARMA (or AR) model.

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The remainder of the thesis is organized as follows. Chapter II contains methods and results of investigating various autoregressive image models. Chapter III deals with the formulation, development, and testing of the twodimensional summation filter. Chapter IV contains the results of applying various AR-generated images to the summation filter and addresses the application of the ARIMA model to real image textures. Chapter V outlines conclusions on the results and applicability of the ARIMA model. Although the ARIMA modeling was not highly successful in reproducing the textures studied here, plausible reasons are given for their failure and conjectures are made about those circumstances where the model would be more successful. Appendix A provides information on the computer algorithms implement the equations governing the processes. Appendices B through G contain derivations of spectral and autocorrelation equations, and the corresponding graphical results, governing the AR processes in Chapter II. Appendices H and I contain graphical results associated with the inverse filter development in Chapter III.

Image data were generated using computer programs written in FORTRAN, compiled using Version 4.5 under the VAX/VMS Version 4.4 operating system. The images were displayed on the COMTAL Vision One/20. The gray level

intensity range of pixel values is 0 (darkest) to 255 (lightest), so the image data generated had to be scaled to that range for display (see Appendix A).

II. THE AUTOREGRESSIVE IMAGE MODEL

A two-dimensional signal (such as an image texture) can be modeled using a two-dimensional AR model with white noise input. The governing equations in the spatial domain are of the following form [Ref. 3:pp. 325-326]:

$$y(n,m) = -\sum_{i=0}^{P-1} \sum_{j=0}^{Q-1} a_{ij}y(n-i,m-j) + w(n,m)$$

$$(2.1)$$

$$(i,j) \neq (0,0)$$

where y(n,m) is a signal representing the generated image texture at pixel location (n,m), a_{ij} is the filter coefficient matrix, and w(n,m) is a two-dimensional white noise signal. The system function corresponding to the filter of Eq. (2.1) is given by

$$Y(z_{1},z_{2}) = \frac{1}{1+a_{10}z_{1}^{-1}+a_{01}z_{2}^{-1}+a_{11}z_{1}^{-1}z_{2}^{-1}+\dots+a_{p-1Q-1}z_{1}^{-(p-1)}z_{2}^{-(p-1)}} \cdot w(z_{1},z_{2})$$

$$(2.2)$$

where z_1 and z_2 are the Z transform variables corresponding to spatial coordinates n and m. Ideal white noise has an autocorrelation function that is an impulse and a flat (constant) power spectrum with magnitude corresponding to the variance of the white noise process [Ref. 4:pp. 22-26]. Therefore determination of the filter coefficients a_{ij} will define the generated image process. Procedures will be

outlined later to estimate the coefficients from real image data. At this point analytical methods will be used to select these coefficients and the resulting images will be studied. Figure 2-1 shows an example of a white noise input image.

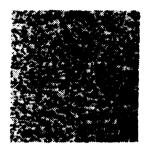


Figure 2-1 White Noise Image

A. ARBITRARILY SELECTED FILTER COEFFICIENTS; P = 2, Q = 2

In order to get an initial idea of what types of images might be generated using a 2×2 AR filter with white noise input, filter coefficients were at first selected arbitrarily, but subject to a stability constraint. The primary constraint on coefficient selection is that of filter stability. Using the DeCarlo-Strintzis Theorem dealing with multidimensional filter stability [Ref. 3:pp. 197-198], alternately setting $z_1 = 1$ and $z_2 = 1$ and determining the location of the pole in the remaining dimension will indicate whether or not the filter is stable. If the

magnitude of the pole in the remaining dimension is less than 1, the filter is stable. Even with this condition, however, there are an infinite number of possible filter coefficient combinations. The additional constraint of $a_{10} = a_{01}$ can be used, and comparisons of results using various values of a_{11} can be made.

Figure 2-2 shows the form and directionality convention used for the autoregressive filter, along with the corresponding difference equation and its Z transform.

$$Y(z_1, z_2) = H(z_1, z_2) \cdot W(z_1, z_2) = \frac{1}{1 + a_{10} z_1^{-1} + a_{01} z_2^{-1} + a_{11} z_1^{-1} z_2^{-1}} \cdot W(z_1, z_2)$$

Figure 2-2 Autoregressive Filter Impulse Response,
Difference Equation, and Z Transform

Although it is difficult to make precise predictions in two dimensions, one can expect that the sign and magnitude of a_{10} or a_{01} would influence the correlation between pixels

in the corresponding directions. For example, a positive value for a_{10} might be expected to yield an image with substantial variation in the n direction (low correlation), particularly if the magnitude of a_{10} is near 1. A negative value for a_{01} with magnitude near 1 might yield an image with lower variations in pixel intensity (high correlation) in the m direction. Since the filter is not necessarily separable (i.e., the denominator of $H(z_1,z_2)$ cannot be factored into the form $D(z_1)\cdot D(z_2)$), conclusions drawn from this line of reasoning may not be completely correct.

Initial attempts at generating images with arbitrarily selected coefficients yielded rather uninteresting results having very little contrast or discernible pattern. Continued experimentation with combinations $a_{10} = a_{01} < 0$ eventually yielded more interesting image textures. Figures 2-3 and 2-4 show the results of using the constraint $a_{10} = a_{01} = -0.35$ and various values of a_{11} for the filter coefficients. For positive values of a_{11} , the images are rather "grainy," with higher magnitudes yielding a somewhat "finer" graininess. There are also some overall intensity differences observed. For the negative values of a₁₁, the results are much more interesting. As the magnitude increases, there is a gradually more noticeable upper left to lower right orientation of the image texture, and the variations from lower left to upper right become smoother as well. Using $a_{10} = a_{01} = -0.38$ and $a_{11} = -0.24$

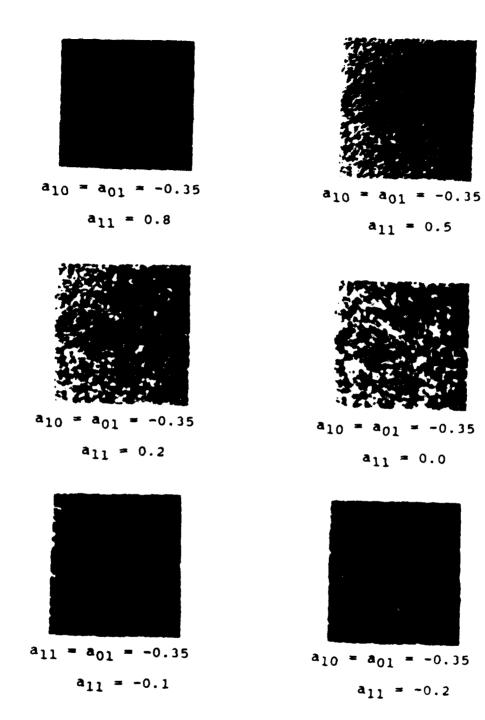


Figure 2-3 Images Generated Using Arbitrarily Selected Filter Coefficients



 $a_{10} = a_{01} = -0.35$ $a_{11} = -0.25$



$$a_{10} = a_{01} = -0.35$$

 $a_{11} = -0.27$



 $a_{11} = -0.3$



 $a_{10} = a_{01} = -0.38$

$$a_{11} = -0.24$$

Figure 2-4 Images Generated Using Arbitrarily Selected Filter Coefficients

yields minor variations in texture pattern and overall image intensity when compared to the previous case.

Using $a_{10} = a_{01} = 0.35$ and varying a_{11} from 0.0 to 0.8 (Figure 2-5), the images obtained deviate very little from the mean intensity value, and possess minor differences in graininess. With these positive coefficients some negative correlation might be expected, and the fact that these images are "grainy" indicates the existence of some negative correlation or high spatial frequency characteristics. On initial examination, however, the low contrast of the generated images tends to obscure the observed graininess.

B. TWO POLE, SEPARABLE AUTOREGRESSIVE MODEL

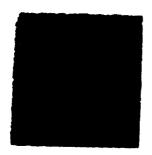
In general, it is difficult to relate the nature or properties of a two-dimensional filter to the precise nature of an image texture that may be generated when white noise is applied to that filter. In order to simplify the effort and to obtain a better understanding of the problem, the case where the filter (and resulting image texture) are separable is considered. For the two pole separable case considered here, the filter transfer function can be factored into expressions in z_1 alone and z_2 alone. The expressions in z_1 and z_2 each have one pole on the real axis in their respective Z domains. Figure 2-6 illustrates the filter structure, the corresponding difference equation, and its Z transform.



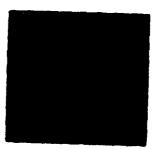
 $a_{10} = a_{01} = 0.35$ $a_{11} = 0.0$



 $a_{10} = a_{01} = 0.35$ $a_{11} = 0.2$



 $a_{10} = a_{01} = 0.35$ $a_{11} = 0.5$



 $a_{10} = a_{01} = 0.35$ $a_{11} = 0.8$

Figure 2-5 Images Generated Using Arbitrarily Selected Filter Coefficients

$$y(n,m) = -\sum_{i=0}^{1} \sum_{j=0}^{1} a_{ij} \cdot y(n-i,m-j) + w(n,m)$$

$$(a_{11} - a_{10} \cdot a_{01})$$

$$(a_{11} - a_{10} \cdot a_{01})$$

$$\begin{split} \mathtt{Y}(\mathbf{z}_{1},\mathbf{z}_{2}) &= \mathtt{H}(\mathbf{z}_{1},\mathbf{z}_{2}) \cdot \mathtt{W}(\mathbf{z}_{1},\mathbf{z}_{2}) \\ &= \frac{1}{1 + \mathtt{a}_{10} \mathbf{z}_{1}^{-1}} \cdot \frac{1}{1 + \mathtt{a}_{01} \mathbf{z}_{2}^{-1}} \cdot \mathtt{W}(\mathbf{z}_{1},\mathbf{z}_{2}) \\ &= \frac{1}{1 + \mathtt{a}_{10} \mathbf{z}_{1}^{-1} + \mathtt{a}_{01} \mathbf{z}_{2}^{-1} + \mathtt{a}_{10} \cdot \mathtt{a}_{01} \mathbf{z}_{1}^{-1} \mathbf{z}_{2}^{-1}} \cdot \mathtt{W}(\mathbf{z}_{1},\mathbf{z}_{2}) \end{split}$$

Figure 2-6 Autoregressive Filter Form, Difference Equation, and Z Transform

Here it is relatively easy to relate stability of the filter to the location of the poles in the z_1 and z_2 planes. Since the quarter plane filter is separable and the components are causal, one-dimensional filter stability theory can be used to state that the poles in each plane must have magnitude less than 1 to ensure filter stability. Figures 2-7 through 2-9 show images resulting from this model for various values of a_{10} and a_{01} . Note that the sign convention used for the filter difference equation and 2 transform results in poles on the negative side of the real axis for positive values of a_{10} or a_{01} , and vice versa-

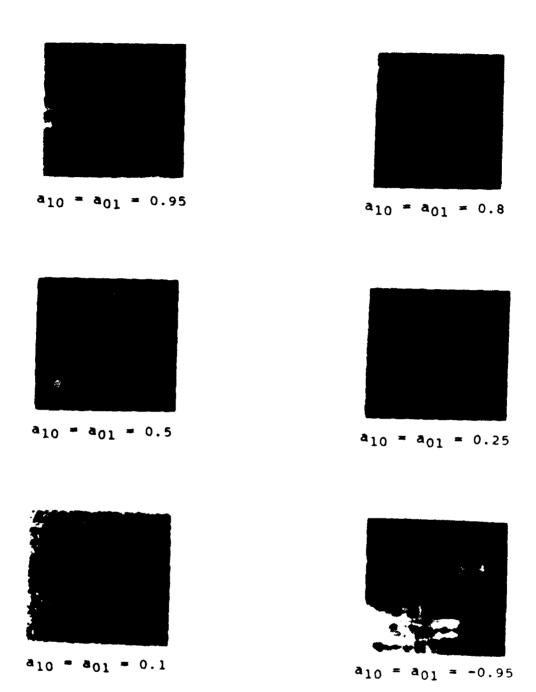


Figure 2-7 Images Generated Using a Two-Pole Autoregressive Model

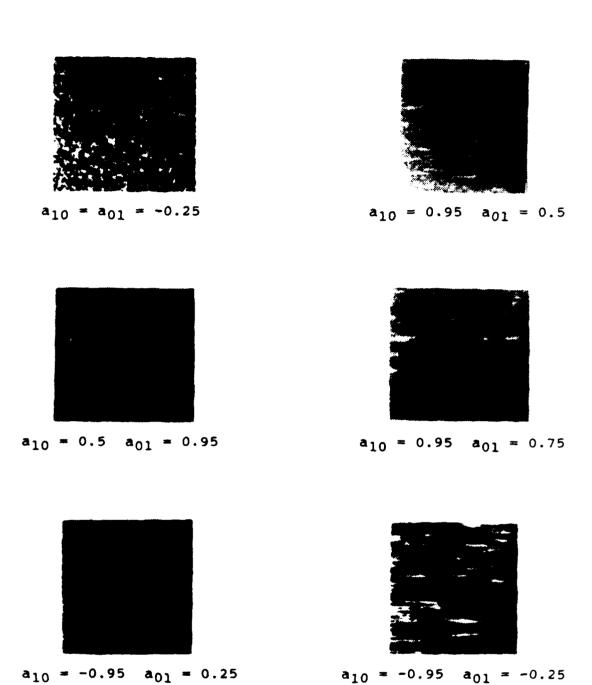
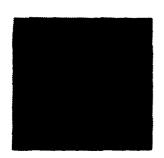
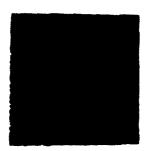


Figure 2-8 Images Generated Using a Two-Pole Autoregressive Model





 $a_{10} = -0.95$ $a_{01} = 0.75$ $a_{10} = 0.95$ $a_{01} = -0.95$



 $a_{10} = 0.25$ $a_{01} = -0.25$

Figure 2-9 Images Generated Using a Two-Pole Autoregressive Model

Careful comparison of the images resulting from various combinations of a_{10} and a_{01} leads to the following observations:

- 1) When poles are located in the same place in the z_1 and z_2 planes on the negative side of the real axis, magnitudes near 1 yield a fine graininess with patchy areas and low overall contrast. As the magnitude of the pole decreases, the graininess becomes more coarse and the result is more like the original white noise input. No directional quality in the image pattern is observed.
- When poles are located in the same place on the positive side of the z_1 and z_2 real axes, a somewhat different result is observed. For magnitudes near 1, an image of patchy light and dark areas results, with differing amounts of correlation between pixels in different areas. Slightly discernible "lines" in both the horizontal and vertical directions are also observed. For lower pole magnitudes on the positive side of the real axis, the decreased effect of the filter on the white noise input is again observed. This result is more like the white noise and has more contrast than the corresponding result using poles on the negative side of the real axis.

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- 3) For pole placements in the z_1 and z_2 planes which are on the negative side of the real axis and are of unequal magnitude, the results have a very fine graininess and low contrast. Some slight directionality is observable in the image patterns, with lower frequency variations evident in the direction corresponding to the pole with smaller magnitude.
- 4) For pole placements in the z_1 and z_2 planes which are on the positive side of the real axis and are of unequal magnitude, much more directionality and variation is observable in the image pattern.
- As the poles are placed on opposite sides of the real axis and are separated by a greater distance, directionality becomes more evident (with higher frequency variations in the direction of the more negative pole). As the pole separation becomes greater and as the pole magnitudes become closer to 1, smoother sinusoidal variations are evident.
- 6) When the pole magnitudes are equal and have opposite sign, the image generated using pole magnitudes close

to 1 exhibits high frequency sinusoidal variations in the direction of the negative pole. The image generated with the lower magnitude poles, as would be expected from the above results, resembled the unfiltered white noise.

In order to explain these image patterns analytically, analysis of the power spectrum and autocorrelation function of this process is useful. Since this model is separable, the analysis can be conducted in each direction separately. The power spectrum is defined by [Ref. 4:pp. 24-34]:

$$S_{\mathbf{Y}}(\omega) = \sigma^{2} |H(e^{\mathbf{j}\omega})|^{2} = \sigma^{2}H(e^{\mathbf{j}\omega})H(e^{-\mathbf{j}\omega}) \qquad (2.3a)$$

where

$$H(e^{j\omega}) = H(z)|_{z=e^{j\omega}}$$
 (2.3b)

and for this case

$$H(z) = \frac{1}{1 + \alpha z^{-1}}$$
 (2.3c)

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Here σ^2 is the magnitude of the white noise power spectrum. We can assume that $\sigma^2 = 1$ with no loss of generality of the results, since σ^2 does not affect the shape of the frequency response.

The autocorrelation function is related to the filter transfer function through the equations [Ref. 5:pp. 391-395]:

$$h(n) = Z^{-1}[H(z)]$$
 (2.4a)

$$y(n) = h(n)*w(n)$$
 (2.4b)

$$R_{\mathbf{Y}}(\ell) = \sigma^{2} \sum_{n=-\infty}^{\infty} h(n) \cdot h(n-\ell)$$
 (2.4c)

Specific forms of the power spectrum and autocorrelation function are given in Appendix B. Since $R_Y(\ell) = R_Y(-\ell)$ [Ref. 5:p. 388], calculating the expression for $R_Y(\ell)$, $\ell < 0$ is not necessary. From Appendix B, the results are:

$$S_{\mathbf{Y}}(\omega) = \frac{1}{1 + 2\alpha \infty s(\omega) + \alpha^2}$$
 (2.5)

$$R_{\mathbf{Y}}(\ell) = \frac{(-\alpha)^{\ell}}{1 - \alpha^2} \qquad \ell \ge 0$$
 (2.6)

Appendix C shows the results of these equations graphically for various values of α . The relationship between the power spectra and their corresponding autocorrelation functions conforms to the expected results from theory (i.e., low frequency spectrum with smooth autocorrelation function, and high frequency power spectrum with rapidly varying autocorrelation function) [Ref. 6:pp. 139-142]. The plots in Appendix C also demonstrate that: 1) For poles on the positive side of the real axis in the Z plane low frequencies predominate and for poles on the negative side of the real axis high frequencies predominate, and 2) Lower

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magnitudes of α result in a broader power spectrum and a wider range of frequencies of significant magnitude. of these observations agree with the image results. For images generated using a more negative pole in a given direction, fine, high frequency graininess is observed in that direction (though the low contrast or low variation about the mean intensity may tend to make this effect less noticeable). When a more positive pole is used, lower frequency variations are more evident in the corresponding direction. As lower magnitudes are used for α in a given direction, more random variations (indicative of a wider range of significant frequency components) are observed in that direction. The form of the autocorrelation function for these cases approaches the autocorrelation function for white noise, i.e., an impulse. Negative poles should yield high frequencies since the negative side of the real axis in the Z plane represents a digital spatial frequency of π , while positive poles in the Z plane correspond to a digital spatial frequency of zero. Note that even when the poles are placed such that a spatial frequency of zero should predominate, there are some low frequency random variations in the resulting images. Since the power spectra of the positive poles all contain some non-zero frequency components (they are not perfect impulses at zero), this characteristic is expected.

C. FOUR POLE, SEPARABLE AUTOREGRESSIVE MODEL

For the cases considered in this section, H(z) again can be factored into expressions in z_1 and z_2 . However here each factor is a 2nd degree polynomial with two poles in the denominator. Figure 2-10 illustrates the filter structure, the applicable difference equation, and the corresponding Z transform.

$$y(n,m) = -\sum_{i=0}^{2} \sum_{j=0}^{2} a_{ij} \cdot y(n-i,m-j) + w(n,m)$$

$$x_{i=0} = 0$$

$$y(n,m) = -\sum_{i=0}^{2} \sum_{j=0}^{2} a_{ij} \cdot y(n-i,m-j) + w(n,m)$$

$$x_{i=0} = 0$$

$$\begin{split} &\mathbf{Y}(\mathbf{z}_{1},\mathbf{z}_{2}) = \mathbf{H}(\mathbf{z}_{1},\mathbf{z}_{2}) \cdot \mathbf{W}(\mathbf{z}_{1},\mathbf{z}_{2}) = \frac{1}{1 + \mathbf{a}_{10} \mathbf{z}_{1}^{-1} + \mathbf{a}_{20} \mathbf{z}_{1}^{-2}} \cdot \frac{1}{1 + \mathbf{a}_{01} \mathbf{z}_{2}^{-1} + \mathbf{a}_{02} \mathbf{z}_{2}^{-2}} \cdot \mathbf{W}(\mathbf{z}_{1},\mathbf{z}_{2}) \\ &= \frac{1}{1 + \mathbf{a}_{10} \mathbf{z}^{-1} + \mathbf{a}_{20} \mathbf{z}_{1}^{-2} + \mathbf{a}_{01} \mathbf{z}_{2}^{-1} + \mathbf{a}_{02} \mathbf{z}_{2}^{-2} + \mathbf{a}_{11} \mathbf{z}_{1}^{-1} \mathbf{z}_{2}^{-1} + \mathbf{a}_{21} \mathbf{z}_{1}^{-2} \mathbf{z}_{2}^{-1} + \mathbf{a}_{12} \mathbf{z}_{1}^{-1} \mathbf{z}_{2}^{-2} + \mathbf{a}_{22} \mathbf{z}_{1}^{-2} \mathbf{z}_{2}^{-2}} \cdot \mathbf{W}(\mathbf{z}_{1},\mathbf{z}_{2}) \end{split}$$

where

$$a_{11} = a_{10} \cdot a_{01}$$
; $a_{21} = a_{10} \cdot a_{02}$; $a_{12} = a_{02} \cdot a_{10}$; $a_{22} = a_{20} \cdot a_{02}$

Figure 2-10 Autoregressive Filter Form, Difference Equation, and Z Transform

Since all of the a_{ij} coefficients are real, the poles must 1) both be on the real axis, or 2) occur in complex conjugate pairs in the z_1 and z_2 planes. Again, pole magnitudes must be less than 1 to ensure filter stability. We will assume here that the poles in each of the factors have equal magnitudes and opposite (or 0 or $\pm \pi$) phase. Letting

 α_1 = magnitude of poles in the z_1 plane

 θ_1 = pole angle (phase) in the z_1 plane

 α_2 = magnitude of poles in the z_2 plane

 θ_2 = pole angle (phase) in the z_2 plane and using Euler's relation, the denominators of $H(z_1)$ and $H(z_2)$ can be expressed as follows:

$$1 + a_{11}z_1^{-1} + a_{20}z_1^{-2} = (1 - \alpha_1 e^{j\theta_1} z_1^{-1}) (1 - \alpha_1 e^{-j\theta_1} z_1^{-1}) = 1 - 2\alpha_1 \infty s(\theta_1) \cdot z_1^{-1} + \alpha_1^2 z_1^{-2}$$

$$1 + a_{01}z_2^{-1} + a_{02}z_2^{-2} = (1 - \alpha_2 e^{j\theta_2} z_2^{-1}) (1 - \alpha_2 e^{-j\theta_2} z_2^{-1}) = 1 - 2\alpha_2 \infty s(\theta_2) z_2^{-1} + \alpha_2^2 z_2^{-2}$$

Hence:

$$a_{10} = -2\alpha_1 \cdot \cos(\theta_1)$$
 $a_{01} = -2\alpha_2 \cdot \cos(\theta_2)$
 $a_{20} = \alpha_1^2$ $a_{02} = \alpha_2^2$

This gives a relationship between the pole magnitude and angle in the Z domain and the filter coefficients in the spatial domain.

1. Complex Conjugate Poles

Figures 2-11 through 2-13 illustrate images generated using this model for various complex conjugate pole combinations in the z_1 and z_2 directions. In comparing each of these image textures in terms of the relative effect of pole positioning in each direction, the following observations can be made:

- 1) For images generated using poles of equal magnitude and angle in both directions, graininess with no directionality to the pattern resulted. Higher pole angles yielded finer (higher frequency) graininess and less contrast. Lower magnitude poles yielded a more random and less structured graininess pattern at the same pole angle.
- Using a pole angle of zero (pole on positive real axis) in one direction and a pole of some non-zero angle in the other direction yielded images with highly directional sinusoidal patterns. The direction corresponding to the pole on the real axis was not totally devoid of variation, but variations were slow, i.e., of very low frequency. The spatial frequency of the sinusoidal pattern can be increased by increasing Large magnitude, high pole angle the pole angle. combinations yielded much cleaner and more structured textures than low magnitude, low pole angle combina-Lower magnitude, high pole angle combinations yielded less structured textures where directionality was evident but the sinusoidal pattern was obscured. Low magnitude, low pole angle combinations yielded very random, non-structured textures of relatively high contrast.
- 3) Using poles in the z_1 and z_2 planes with the same magnitude but different pole angles resulted in some directionality if there was a sufficiently large magnitude and difference in the pole angles. As observed earlier, the direction with the higher pole angle had the higher spatial frequency. Large pole magnitudes (close to 1) resulted in more structured but rather low-contrast images (the low contrast seemed to obscure the high frequency nature of the pattern somewhat). Pole angles in z_1 and z_2 that were of close value made it difficult to detect the higher frequency (higher pole angle) direction. Low



 z_1 0.8e $\frac{1}{3}$ z_2 0.8e



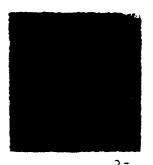
 z_1 0.9e $\frac{1}{3}$ z_2 0.9e $\frac{2\pi}{3}$



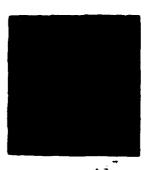
 $z_1 = 0.9e^{\frac{z}{10}}$ $z_2 = 0.9e^{\frac{z}{3}}$



 z_1 0.7e $\frac{7\frac{1}{3}}{2}$ z_2 0.7e



 z_1 0.6e $\frac{1}{2}$ 0.6e $\frac{1}{3}$

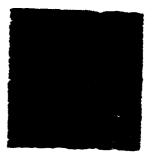


 z_1 0.9e^{$\frac{1}{3}$} z_2 0.9e^{$\frac{1}{3}$}

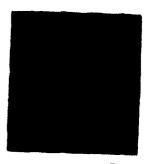
Figure 2-11 Images Generated Using a Four Pole Autoregressive Model (Poles Listed Below Image)



 z_1 0.9e $\frac{100}{2}$ z_2 0.9e



 z_1 0.6e $\frac{10}{27}$ z_2 0.6e $\frac{10}{3}$



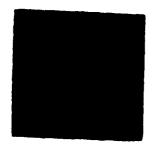
 z_1 0.9e $\frac{\pm j\frac{\pi}{6}}{5}$ z_2 0.9e $\frac{\pm j\frac{\pi}{6}}{6}$



 z_1 0.9e $3\frac{1}{5}$ z_2 0.9e



 z_1 0.6e $\frac{10}{6}$ z_2 0.6e



 z_1 0.9e $\frac{15}{12}$ z_2 0.9e $\frac{15}{12}$

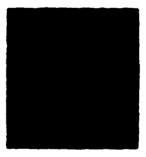
Figure 2-12 Images Generated Using a Four Pole Autoregressive Model (Poles Listed Below Image)



$$z_1$$
 0.6e $\frac{15}{5}$ $\frac{5}{6}$ z_2 0.6e



$$z_1$$
 0.5e $\frac{1}{6}$ z_2 0.5e



$$z_1 \quad 0.9e^{\frac{z}{1}}$$

$$z_2 \quad 0.8e^{\frac{z}{2}}$$



 z_1 0.6e $\frac{1}{12}$ z_2 0.9e

Figure 2-13 Images Generated Using a Four Pole Autoregressive Model (Poles Listed Below Image)

magnitudes basically negated the pole angle effects and yielded a very random, unstructured, high contrast texture.

4) Where pole magnitudes were close and pole angles were different in the z_1 and z_2 planes, some directionality in the texture was observed. Again, the high pole angle direction yielded the highest frequency. When the pole angles in both Z domains had similar values and the magnitudes of the poles differed, graininess with little or no discernible directionality or structure resulted.

These observations are consistent with classical pole-zero frequency response analysis [Ref. 7:pp. 323-331]. There is a direct relationship between pole angle and spatial frequency in a given direction, and the magnitude of the poles affects the amount of structure and definition of the sinusoidal pattern of a given frequency in a given Higher magnitude poles result in a narrower direction. bandwidth of the filter and yield more structure and sinusoidal pattern definition. Low pole magnitudes give the filter wider bandwidth and yield images with less structure and definition and more randomness in a given direction. While directional dependencies are evident given pole magnitude and angle in a given direction, it does not appear that a pattern in one direction is totally independent of a pattern in the other direction. This would be expected, even though the model is separable, due to the cross terms in the filter structure.

Filter power spectrum and autocorrelation analysis can be conducted in this case, as in the case of the two pole model. The derivations for $S_V(\omega)$ and $R_V(\lambda)$ are

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somewhat more involved, and are given in Appendix D. The resulting expressions for $S_y(\omega)$ and $R_y(\hat{z})$ from Appendix D are:

$$s_{\mathbf{y}}(\omega) = \frac{1}{1+\alpha^4+2(\alpha^2-2[\alpha^3-\alpha]\cos(\theta)\cos(\omega)+\alpha^2(\cos(2\theta)+\cos(2\omega)))}$$
 (2.7)

$$\mathbf{R}_{\mathbf{Y}}(\ell) = \frac{\alpha^2}{2\sin^2\theta} \left(\frac{\cos(\ell\theta)}{1-\alpha^2} - \frac{\cos((2+\ell)\theta) - \alpha^2\cos(\ell\theta)}{1+\alpha^4 - 2\alpha^2\cos(2\theta)}\right) \quad (\ell \ge 0)$$
 (2.8)

The plots of these functions for the various pole magnitudes and angles used are given in Appendix E. θ = 0 case is equivalent to having 2 poles on the real axis at a given magnitude in the Z plane. As would be expected, the power spectrum for each model showed higher magnitudes at digital frequencies close to the pole angle. Higher pole magnitudes yielded sharper, more well-defined power spectrum magnitudes at the given frequency, and lower pole magnitudes yielded less well defined more spread-out power spectra. Low pole magnitudes almost completely obliterated evidence of low frequency power spectrum components, and degraded its definition and sharpness at higher frequencies. This corresponds to the observed results in the image textures generated. The autocorrelation functions also reflect the appropriate relationship to the power spectra as outlined in the discussion for the two pole case, i.e., greater variation in the autocorrelation function indicates greater variation between pixels a given distance apart, which in turn implies higher spatial frequencies.

2. Two Real Poles

Rather than using complex conjugate pole locations to obtain real filter coefficient values, two poles on the real axis may also be used for a given direction. They may be placed at different locations on the real axis, or they may be placed together. The latter situation is equivalent to the $\theta=0$ (or $\theta=\pm\pi$ if placed on the negative real axis) case, as mentioned above. For the two real pole case, the relevant equations are:

$$H(z) = \frac{1}{(1-\alpha_{a}z^{-1})(1-\alpha_{b}z^{-1})} = \frac{1}{1-(\alpha_{a}+\alpha_{b})z^{-1}+\alpha_{a}\alpha_{b}z^{-2}}$$
$$= \frac{1}{1+a_{01}z^{-1}+a_{02}z^{-2}}$$

where

$$\mathbf{a_{01}} = -(\alpha_{\mathbf{a}} + \alpha_{\mathbf{b}})$$

$$a_{02} = x a^{*x}b$$

For these experiments a transfer function of the complex conjugate pole form was used for the z_1 direction, with $z_1=0.9$ and $z_1=0$. For the z_2 direction a transfer function with two poles on the real axis was used. The

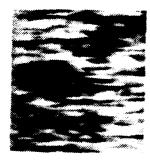
image textures that result for various values of α and α and α are given in Figures 2-14 and 2-15. Some observations can be made about these results:

- 1) With poles placed at the same value on the z_2 real axis, rather unstructured, low frequency variations are observed in the z_2 direction. The more positive poles result in very slow variation in the image texture, while the lower magnitude positive poles show more variation in the z_2 direction. When the poles are placed in the same location on the negative side of the real axis, a low contrast image with some noticeable high frequency variations results.
- 2) As the poles are moved farther apart on the z_2 real axis, high frequency variations with increasing structure and oscillatory form are evidenced in the z_2 direction.
- 3) When poles with equal magnitude and opposite sign are used, fairly structured high frequency variations are evidenced in certain areas of the image, while low frequency variations are evident in other areas in the z₂ direction. Higher magnitude poles yield more discernible, structured variations, while lower magnitude poles of opposite sign yield discernible but non-oscillatory high frequency variations in certain areas of the image.

Of particular interest is the fact that two poles placed at the same value on the negative real axis in the z_2 plane yielded some high frequency variations. This is in keeping with the fact that values on the negative real axis correspond to a pole angle (and corresponding digital frequency) of $\theta=\pi$. The presence of poles on the negative side of the real axis of the z_2 p are seems to give rise to the high frequency variations with gradually more structure and oscillatory appearance as the pole is moved to the left (more negative).



 $\alpha_{a} = 0.9 \quad \alpha_{b} = 0.9$



 $\alpha_{a} = 0.5 \quad \alpha_{b} = 0.5$



 $\alpha_{a} = -0.9 \quad \alpha_{b} = -0.9$



 $\alpha_{a} = 0.9 \quad \alpha_{b} = 0.5$

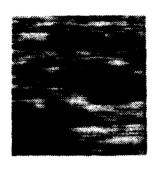


 $a_a = 0.9 \quad a_b = 0.0$



 $\alpha_{a} = 0.9 \quad \alpha_{b} = -0.2$

Figure 2-14 Images Generated Using a Four Pole (Two Real Poles) Autoregressive Model



 $\alpha_a = 0.8 \quad \alpha_b = -0.9$



 $\alpha_{a} = 0.9 \quad \alpha_{b} = -0.9$



 $\alpha_{a} = 0.5 \quad \alpha_{b} = -0.5$

Figure 2-15 Images Generated Using a Four Pole (Two Real Poles) Autoregressive Model

The expressions for the power spectrum and autocorrelation function for the random process produced by driving the filter of Figure 2-10 with white noise are derived in Appendix F. It is shown there that the power spectral density and the autocorrelation function are given by:

$$S_{\mathbf{y}}(\omega) = \frac{1}{1-2(\alpha_{\mathbf{a}}+\alpha_{\mathbf{b}}+\alpha_{\mathbf{a}}^{2}\alpha_{\mathbf{b}}+\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}^{2})\cos(\omega)+2\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}\cos(2\omega)+\alpha_{\mathbf{a}}^{2}} + 2\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}+\alpha_{\mathbf{b}}^{2}+\alpha_{\mathbf{a}}^{2}\alpha_{\mathbf{b}}^{2}}$$
(2.9)

$$R_{\mathbf{Y}}(\ell) = \frac{1}{(\alpha_{\mathbf{a}} - \alpha_{\mathbf{b}})^{2}} \left[\frac{\alpha_{\mathbf{a}}^{2+\ell}}{1 - \alpha_{\mathbf{a}}^{2}} - \frac{\alpha_{\mathbf{a}}^{\ell+1} \alpha_{\mathbf{b}} + \alpha_{\mathbf{a}} \alpha_{\mathbf{b}}^{\ell+1}}{1 - \alpha_{\mathbf{a}}^{2} \alpha_{\mathbf{b}}} + \frac{\alpha_{\mathbf{b}}^{2+\ell}}{1 - \alpha_{\mathbf{b}}^{2}} \right]$$
(2.10)

Plots of these functions for the various values used in this section are given in Appendix G. The power spectrum results are consistent with the observed image spatial frequency characteristics. Both low and high frequency components were contained in some of the power spectra, and were manifested in the corresponding images as both low and high frequency variations in the z_2 direction. The nature of the autocorrelation functions related to the power spectra that contained low and high frequency components was interesting. Autocorrelation functions with much variation but all positive values, rather than the equal magnitude

positive and negative values evidenced in earlier results, seems to reflect the higher level of correlation related to the low frequency (smoother variations) aspect of the image texture variations.

D. IMAGE TEXTURE ROTATION TRANSFORMATION

If an image signal $x(n_1,n_2)$ consists of a rotated version of another image $w(m_1,m_2)$ such that $n_1=Im_1+Jm_2$ and $n_2=Km_1+Lm_2$, where I, J, K, and L are integers and $IL-KJ\neq 0$, then the Z transform $X(z_1,z_2)$ is given by $W(z_1^I,z_2^K,z_1^J,z_2^L)$ [Ref. 3:p. 182]. A 45° rotation corresponds to I = 1, K = 1, J = 1, L = -1. If we use the four pole separable result for $H(z_1,z_2)$, as shown in Figure 2-10, and apply the above transformation $(z_1+z_1^1z_2^1,z_2+z_1^1z_2^{-1})$, we find after simplification:

$$H_{R}(z_{1}z_{2}) = \frac{1}{1+a_{01}z_{1}^{-1}z_{2}^{-1}+a_{02}z_{1}^{-2}z_{2}^{2}+a_{10}z_{1}^{-1}z_{2}^{-1}+a_{10}a_{01}z_{1}^{-2}+a_{10}a_{02}z_{1}^{-3}z_{2}^{-1}} + a_{20}z_{1}^{-2}z_{2}^{-2}+a_{20}a_{01}z_{1}^{-3}z_{2}^{-1}+a_{20}a_{02}z_{1}^{-4}}$$
(2.11)

Notice that this transfer function is not separable but consists of a rotated version of a separable filter. Figure 2-16 illustrates the support of the denominator polynomial for this filter. It has the form of a non-symmetric halfplane infinite impulse response (IIR) filter, so it is recursively computable.

0	0	a ₂₀	0]
0	a ₂₀ a ₀₁	0	a ₁₀	
a ₂₀ a ₀₂	0	a ₁₀ a ₀₁	0	1
0	a ₁₀ a ₀₂	0	a ₀₁	0
0	0	a ₀₂	0	0

Figure 2-16 Rotation Transformation Filter Form

The application of this filter, using filter coefficients of the four pole separable filter with poles at $z_1 \rightarrow 0.9e^{\frac{-j\pi}{2}}$, $z_2 \rightarrow 0.9e^{\pm j0}$ in the original separable filter yielded the result shown in Figure 2-17.



Figure 2-17 Result of Rotation Transformation

E. SUMMARY

Autoregressive models can produce a variety of image textures. For general two-dimensional models, the system functions are generally not factorable and singularities occur on surfaces, not at isolated points. For these reasons it is difficult to design two-dimensional filters for images and predict the resulting character of the images. Indeed, even to ensure stability of the filter is not trivial. As a result we concentrated here on separable forms, which by their nature are much easier to analyze. Certain types of texture patterns using various separable autoregressive models can be predicted based on filter pole placement in the \mathbf{z}_1 and \mathbf{z}_2 planes. Arbitrary or random selection of filter coefficients can yield interesting but generally unpredictable results. Obviously, an infinite number of variations on the models above could be attempted. Ultimately, the anticipated utility of the textures generated will guide the process of model and parameter selection.

III. IMPLEMENTATION OF AN FIR SUMMATION FILTER IN TWO DIMENSIONS

To implement the 2-D ARIMA model, the inverse of a filter representing a suitable difference operator is needed. One possible 2-D difference operator is the Laplacian, which has the impulse response shown in Figure 3-1 [Ref. 8:pp. 212-213].

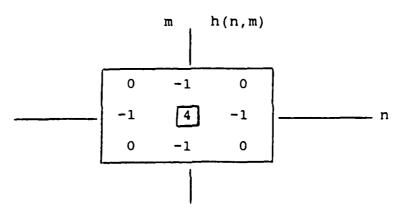


Figure 3-1 Laplacian Impulse Response

Its implementation involves convolving it with an image and is represented by the following difference equation:

$$y(n,m) = \sum_{i=-1}^{1} \sum_{j=-1}^{1} b_{ij}x(n-i,m-j)$$
 (3.1)

where

x(n,m) is the image input signal, and b_{ij} is the filter coefficient matrix $(b_{ij} = h(i,j))$

In the Z-transform domain this can be written as:

$$Y(z_1, z_2) = H(z_1, z_2) \cdot X(z_1, z_2)$$

= $(4 - z^{-1} - z_1 - z^{-1} - z_2) \cdot X(z_1, z_2)$

In areas of an image where adjacent pixels have similar gray levels (low frequency, homogeneous areas), the result of this operator will be approximately zero. Where significant or sharp differences in gray levels between adjacent pixels exist, the result of this operation will be farther from zero. Thus the Laplacian difference operator is sometimes used as an "edge detector."

The problem addressed in this chapter is constructing the inverse of the operator. In the Z domain, the expression for the inverse would be [Ref. 4:p. 36]:

$$H^{-1}(z_1,z_2) = \frac{1}{H(z_1,z_2)} = \frac{1}{4-z_1^{-1}-z_1-z_2^{-1}-z_2}$$

which has an expansion as an infinite series of positive and negative powers of z_1 and z_2 . That is, considering this expression as a problem in long division, the result of such division would be an expression of the form:

$$H^{-1}(z_1,z_2) = \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} a_{ij} z_1^{-i} z_2^{-j}$$

where

 $a_{ij} = coefficient values of <math>z_1^i z_2^j$ resulting from the long division

Note that if

$$\frac{1}{4 - z_1^{-1} - z_1 - z_2^{-1} - z_2} = \int_{1 = -\infty}^{\infty} \int_{j = -\infty}^{\infty} a_{ij} z_1^{-i} z_2^{-j}$$

then cross multiplication would yield:

$$(4 - z^{-1} - z_1 - z^{-2} - z_2) \cdot \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} a_{ij} z^{-i} z^{-j} = 1$$
 (3.3)

The double summation expression in z_1 and z_2 will be truncated and considered to be an FIR filter with finite support and coefficients a_{ij} . This approximates the desired inverse filter. In particular, we will use the following constraints:

1) Choose the limits of summation to be equal in both directions, i.e.,

$$\sum_{i=-L}^{L} \sum_{j=-L}^{L} a_{ij} z_{1}^{-i} z_{2}^{-j}$$

This results in a "square" region of support for the filter (all values outside assumed zero).

2) Force the values for the filter coefficients to be symmetric, i.e., $a_{ij} = a_{-ij} = a_{i-j} = a_{-i-j} = a_{ji} = a_{-j-i} = a_{-j-i}$.

Using these constraints and implementing the cross multiplication equation (3.3) will result in an expression in z_1 and z_2 , with each combination of the $\bar{z}_1^i z_2^j$ terms having a coefficient whose form is a summation of terms in a_{ij}

where the coefficients of a_{ij} are either 4 or -1. The coefficient of the $z_1^0 z_2^0$ term must equal 1 and the coefficient of any other $z_1^{-i} z_2^{-j}$ term must equal zero to satisfy equation (3.3).

As an example, let L = 1. Equation (3.3) can then be expressed as:

$$(4 - z^{-1} - z_1 - z^{-1} - z_2) \cdot \frac{1}{2} \int_{i=-1}^{1} a_{ij} z_1^{-i} z_2^{-j}$$

$$= 1 + 0 \cdot z_1^1 + 0 \cdot z_2^1 + 0 \cdot z_1^{1} z_2^1 + \dots$$

Performing the double summation yields

$$\frac{1}{2} \quad \int_{i=-1}^{1} a_{ij} z_{1}^{-i} z_{2}^{-j} = a_{-1-1} z_{1}^{1} z_{2}^{1} + a_{-10} z_{1}^{1} z_{2}^{0} + a_{-11} z_{1}^{1} z_{2}^{-1} + a_{00} z_{1}^{0} z_{2}^{0} + a_{0-1} z_{1}^{0} z_{2}^{1} + a_{1-1} z_{1}^{-1} z_{2}^{1} + a_{10} z_{1}^{-1} + a_{01} z_{2}^{-1} + a_{11} z_{1}^{-1} z_{2}^{-1}$$

Performing the cross multiplication would yield 45 different terms in various combinations of $z_1^{-i}z_2^{-j}$. Combining these terms to find the coefficient expression for each $z_1^{-i}z_2^{-j}$ term soon becomes rather tedious and impractical for even moderate values of L. An alternative way to proceed is to choose a $z_1^a z_2^b$ term on the right hand side of the equation, and for each term in the expression $4-z_1^{-1}-z_1-z_2^{-1}-z_2$, determine what values of i and j are required so that when each term is multiplied by $a_{ij}z_1^{-i}z_2^{-j}$, it will result in an

expression in the chosen $z_1^a z_2^b$ term on the right hand side of the equation. Choosing $z_1^0 z_2^0$ (= 1) on the right hand side of the equation, we have:

So the coefficient for z_{12}^0 is simply a summation of the c terms obtained above, i.e.,

$$(4a_{00} - a_{10} - a_{10} - a_{0-1} - a_{01}) z_1^0 z_2^0$$

This entire expression must equal 1 to satisfy (3.3), and since $z_1^0 z_2^0 = 1$, $4a_{00} - a_{-10} - a_{10} - a_{0-1} - a_{01} = 1$ also.

Using the same method for the $z \, {}^1 z \, {}^1$ term on the right hand side yields:

$$4 \cdot a_{ij} z_{1}^{-i} z_{2}^{-j} = c z_{1}^{1} z_{2}^{1} \text{ when } i = -1 \text{ and } j = -1; \text{ so } c = 4 a_{-1-1}$$

$$-z_{1}^{-1} \cdot a_{ij} z_{1}^{-i} z_{2}^{-j} = c z_{1}^{1} z_{1}^{1} \text{ when } i = -2 \text{ and } j = -1; \text{ so } c = -a_{-2-1}$$

$$-z_{1} \cdot a_{ij} z_{1}^{-i} z_{2}^{-j} = c z_{1}^{1} z_{1}^{1} \text{ when } i = 0 \text{ and } j = -1; \text{ so } c = -a_{0-1}$$

$$-z_{1}^{-1} \cdot a_{ij} z_{1}^{-i} z_{2}^{-j} = c z_{1}^{1} z_{1}^{1} \text{ when } i = -1 \text{ and } j = -2; \text{ so } c = -a_{-1-2}$$

$$-z_{2} \cdot a_{ij} z_{1}^{-i} z_{2}^{-j} = c z_{1}^{1} z_{1}^{1} \text{ when } i = -1 \text{ and } j = 0; \text{ so } c = -a_{-10}$$

The resulting term in $z_1^1 z_2^1$ is:

$$(4a_{-1-1} - a_{-2-1} - a_{0-1} - a_{-1-2} - a_{-10}) z_1^1 z_2^1$$

This expression must equal zero, since there is no $z_1^1 z_2^1$ term on the right side of (3.3), so $4a_{-1-1} - a_{-2-1} - a_{0-1} - a_{-1-2} - a_{-10} = 0$.

This procedure can be extended to any number of $z_1^{-1}z_1^{-1}$ terms. When this is done, the resulting expressions for the coefficients of $z_1^{-i}z_2^{-j}$ can be formed into a set of simultaneous equations in order to solve for the $a_{\dot{1}\dot{1}}$ coefficient values. However, due to the symmetry condition imposed above, some of the equations for the coefficients of the $z_1^{-i}z_2^{-j}$ terms are linearly dependent. For values of i and j that yield unique or distinct values for a_{ij} , the resulting $z_1^{-i}z_2^{-j}$ coefficient expressions are linearly independent. For example, the coefficient expression for the $z_1^{-1}z_2^0$ term is linearly independent of the coefficient expression for the $z_1^{-1}z_2^{-0}$ term, since $a_{10} \neq a_{11}$. But the coefficient expression for the $z_1^1 z_2^0$ term is linearly dependent on the coefficient expression for the $z_{1}^{0}z_{2}^{-1}$ term, since $a_{10} = a_{01}$. Using only the linearly independent equations for a given filter size yields a set of p equations in p unknowns, where p is the number of unique and distinct a_{ij} values in the inverse filter. The value of p is related to the size of the desired inverse filter. If the size of the filter is N · N,

the number of unique a_{ij} values using the symmetry of constraint above is:

$$\mathbf{p} = (\frac{N-1}{2} + 1) + (\frac{N-1}{2}) + (\frac{N-1}{2} - 1) + \dots + 1$$
 (3.4a)

$$= \frac{(N+1)(N+3)}{4} \tag{3.4b}$$

For example, with N = 7, the unique a_{ij} values in a 7 · 7 inverse filter can be represented by a_{33} , a_{32} , a_{31} , a_{30} , a_{22} , a_{21} , a_{20} , a_{11} , a_{10} , a_{00} . Though there are 49 elements in a 7 · 7 filter, all of them are equal to one of these values listed, due to symmetry. Obviously, N is constrained to be odd for a square filter with a unique element a_{00} in the center.

The solution of the resulting p equations yields the values for the p filter elements or coefficients. This defines the FIR approximation to the inverse filter. It is only an approximation due to the finite size constraint imposed, and it might be expected that the larger the filter size, the better the approximation.

An algorithmic procedure for obtaining the a_{ij} coefficients is outlined below. An example follows.

- Determine the desired size of the inverse filter.
- 2) For each combination of (positive) i,j values corresponding to a <u>unique</u> a_{ij} filter coefficient, identify the five term summation equation associated with each $z_{1}^{-1}z_{2}^{-j}$ term.

- 3) Combine equal a_{ij} values and develop a matrix of coefficients for the a_{ij} values. Let this matrix be A.
- 4) Denoting the vector of unique aij values as a, the set of simultaneous equations in matrix form is:

$$\overline{A}\overline{a}^{\mathrm{T}} = \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$
 (3.5)

where

$$\bar{a} = [a_{00} \ a_{10} \ a_{11} \ a_{20} \ a_{21} \ a_{22} \dots a_{N-1} \frac{1}{2}]$$

The top row of \overline{A} corresponds to the summation of ajj terms that represents the coefficient of the $z_1^0 z_2^0$ term.

5) Solve (3.5) for \overline{a} .

An example of this procedure is appropriate at this point. For an inverse filter of size $N \times N$:

Step 1

Let N = 5 (therefore L = 2)

Step 2

The coefficients corresponding to each unique $z^{-\dot{1}}z^{-\dot{j}}$ term are:

$$z_1^0 z_2^0 \rightarrow 4a_{00} - a_{10} - a_{-10} - a_{01} - a_{0-1}$$

 $z_1^{-1} z_2^0 \rightarrow 4a_{10} - a_{20} - a_{00} - a_{11} - a_{1-1}$
 $z_1^{-1} z_2^{-1} \rightarrow 4a_{11} - a_{21} - a_{01} - a_{12} - a_{10}$

$$z_{1}^{-2}z_{2}^{0} \rightarrow 4a_{20} - a_{30} - a_{10} - a_{21} - a_{2-1}$$
 $z_{1}^{-2}z_{2}^{-1} \rightarrow 4a_{21} - a_{31} - a_{11} - a_{20} - a_{2-2}$
 $z_{1}^{-2}z_{2}^{-2} \rightarrow 4a_{22} - a_{32} - a_{12} - a_{23} - a_{21}$

Step 3

Combining equal terms in Step 2 and expressing the coefficients of a_{ij} in matrix form yields an \overline{A} matrix of:

$$\overline{A} = \begin{bmatrix}
4 & -4 & 0 & 0 & 0 & 0 \\
-1 & 4 & -2 & -1 & 0 & 0 \\
0 & -2 & 4 & 0 & -2 & 0 \\
0 & -1 & 0 & 4 & -2 & 0 \\
0 & 0 & -1 & -1 & 4 & -1 \\
0 & 0 & 0 & 0 & -2 & 4
\end{bmatrix}$$

Step 4

With \overline{A} given in Step 3, the \overline{a} vector for (3.5) is

$$\bar{a} = [a_{00} \ a_{10} \ a_{11} \ a_{20} \ a_{21} \ a_{22}]$$

Step 5

Solving (3.5) for \overline{a} involves inverting \overline{A} and multiplying it by [1 0 0 0 0]^T. Thus:

$$\overline{\mathbf{a}} = \overline{\mathbf{A}}^{-1} \cdot [1 \quad 0 \quad 0 \quad 0 \quad 0]^{\mathrm{T}}$$

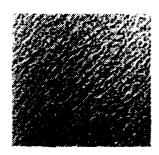
Appendix H illustrates the forms of the resulting inverse filters of various sizes, as well as the normalized and unnormalized filter cross sections.

One way to validate the resulting inverse filter is to convolve it with the original Laplacian difference operator. The result should approximate an impulse at the origin. Appendix I shows the results of this convolution using 3×3 , 5×5 , 7×7 , 9×9 , 15×15 , and 21×21 size inverse filters. It is seen there that as the size of the filter gets larger, it becomes a better approximation to the true inverse and the convolution looks more like an impulse.

To test the application of this filter on an actual image, a test image was filtered using the Laplacian difference operator. Then the resulting image data were filtered again using various size inverse filters. The results are shown in Figures 3-2 and 3-3. Note that the image resulting from Laplacian FIR filtering seems more stationary than the test image, which was one of the desired results. Inverse filtering of that result yields images that are progressively more similar to the original test image as the size of the inverse filter increases. However, a rather large inverse filter is needed to accurately reproduce the image. The result of the 21 × 21 inverse filter is quite similar to the test image, with some loss of contrast or darkness in certain areas, but with essentially the same pattern. The effect of the size limitation of the



Test image



Laplacian filtered test image



 3×3 inverse filter



5 × 5 inverse filter

Figure 3-2 Results of Filtering Test Image with Laplacian FIR Filter and its Inverse



7 × 7 inverse filter



 9×9 inverse filter



15 × 15 inverse filter



21 × 21 inverse filter

Figure 3-3 Results of Filtering Test Image with Laplacian FIR Filter and its Inverse

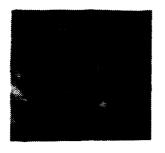
inverse filter, as manifested in the convolution of the Laplacian and its inverse above, would seem to explain the lack of perfect test image reproduction. Larger inverse filter sizes could be tried, but large inverse filter sizes relative to image size would result in a significant portion around the edge of the image having only a part of the filter applied to it. This would adversely affect the overall quality of image reproduction.

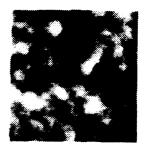
IV. APPLICATION OF THE ARIMA MODEL TO IMAGE TEXTURES

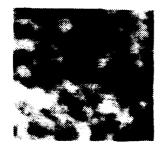
As outlined in Chapter I, the utility of the ARIMA model centers around the fact that a difference operator applied to an image texture may improve the stationarity of the image statistical characteristics. A stationary image texture is required for accurate modeling by autoregressive techniques, and it was hoped that application of the autoregressively generated texture to an approximate inverse of the difference operator may yield a more accurate or recognizable representation of the original nonstationary image, as compared to a purely autoregressively generated version.

A. APPLICATION OF LAPLACIAN INVERSE FILTER TO AUTO-REGRESSIVELY GENERATED IMAGES

As an initial examination of the effects of the inverse filter developed in Chapter III on image textures, selected images generated in Chapter II were input to the 21 x 21 version of that filter. Figures 4-1 through 4-3 illustrate the results. All attempts resulted in a blurred or smoothed version of the original image. Since the inverse of a difference operation is a summation or "integration" operation, and since integration operations can be expected to smooth or blur (low pass filter) signals [Ref. 8:pp. 136not surprising. 154], the results are However,







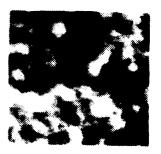


Figure 4-1 Images from Figure 2-7 (top 4) Applied to Summation Filter









Figure 4-2 Images from Eleme 2-11 (top 4) Applied to Summation Elemen









Figure 4-3 Images from Figure 2-14 (top 4) Applied to Summation Filter

except to the extent that blurring is useful or desirable, applying the summation filter to image signals that are not based on the application of the corresponding difference filter to that signal seems to be of little utility.

In the remainder of this chapter we consider application of the summation filter to regenerate actual image textures.

B. AUTOREGRESSIVE FILTER PARAMETER ESTIMATION PROCEDURES

The first step in testing the ARIMA model is to estimate the autoregressive, quarter plane filter parameters required to model the real image textures and the signal resulting from application of the Laplacian operator to those images. For a zero-mean signal, these model parameters are found by solving a set of Normal equations. In these equations the white noise covariance is referred to as the prediction error covariance. The Normal equations can be expressed as

$$\vec{R} \qquad \begin{bmatrix} \vec{a}_0 \\ \vec{a}_1 \\ \vdots \\ \vec{a}_{p-1} \end{bmatrix} = \begin{bmatrix} \vec{s} \\ 0 \\ \vdots \\ 0 \end{bmatrix} \tag{4.1}$$

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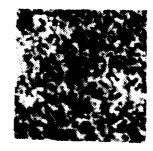
where the R matrix is the correlation matrix for the signal (block Toeplitz with Toeplitz blocks), the a vector consists of appropriately ordered filter coefficient vectors, and s is a vector containing the prediction error covariance as

the first and only nonzero element. Here $\bar{a_i} = [a_{i0} \ a_{i1}]$ $a_{i2} \dots a_{i \ O-1}]^T$ and $\bar{s} = [\sigma^2 \ 0 \ 0 \dots 0]^T$.

Calculating the correlation matrix and prediction error covariance from the image signals, and solving (4.1) for the a vectors, provides all the parameters needed for the 2-D AR model. The multichannel form of the Levinson recursion can be used to solve these equations more efficiently [Ref. 2:p. 454].

C. APPLICATION TO REAL IMAGE TEXTURES

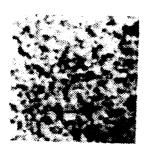
Actual image textures used here are from the image data base at the University of Southern California's Signal and Image Processing Institute [Ref. 9:pp. 13-14]. The images selected from this data base are contained in a book by Brodatz [Ref. 10]. Portions of the images of size 128 × 128 pixels were obtained and used as a basis for processing. Filter coefficients were calculated for the real image textures shown in Figures 4-4 and 4-5. Various filter sizes were tried to determine which yielded the best results in generating a particular image, and a quarter-plane filter size of 4 × 4 was selected. Results of autoregressive filtering of white noise using the appropriate calculated coefficients to model each texture are given in Figures 4-6 and 4-7. Images generated by applying the Laplacian difference operator to the real images are shown in Figures 4-8 and 4-9. Autoregressive generation of these images using



Grass



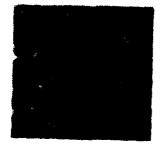
Bark



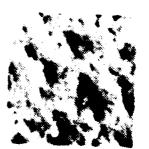
Sand



Water



Bubbles



Wall

Figure 4-4 Actual Image Textures



Sand



Gravel



Grass

Figure 4-5 Actual Image Textures

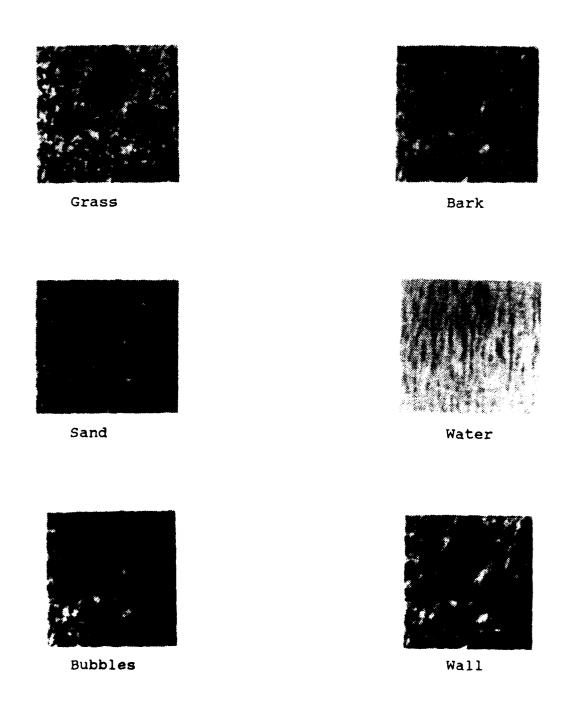
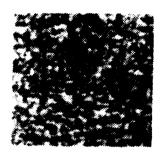


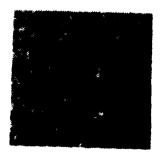
Figure 4-6 Image Textures Generated Using an AR Model



Sand



Gravel



Grass

Figure 4-7 Image Textures Generated Using an AR Model

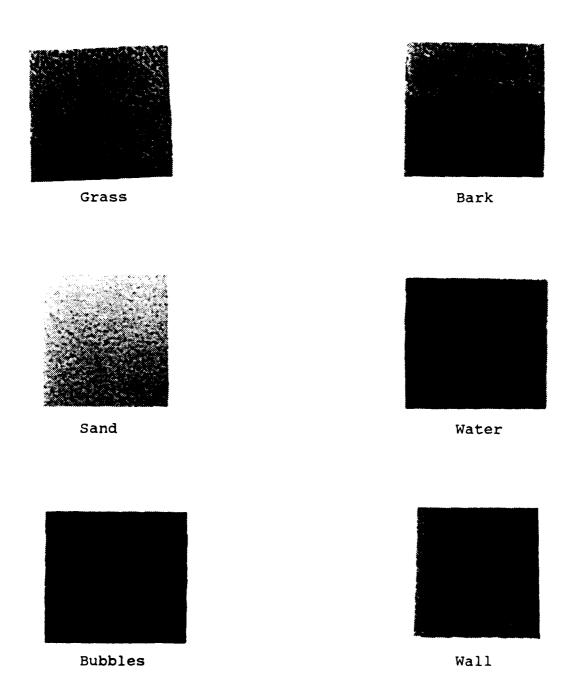
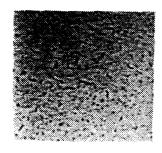
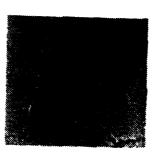


Figure 4-8 Actual Images After Laplacian Filtering



Sand



Gravel



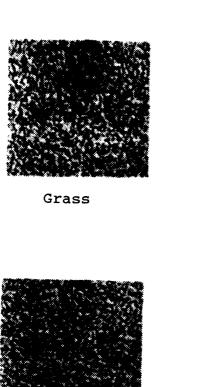
Figure 4-9 Actual Images After Laplacian Filtering

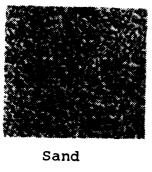
filters with the corresponding calculated coefficients are given in Figures 4-10 and 4-11. Finally, the application of the signal represented by the images in Figures 4-10 and 4-11 (without the 0-255 scaling reflected in these figures) to the 21×21 inverse filter described in Chapter III yields the images shown in Figures 4-12 and 4-13. Comparison of all of the above results yields the following observations:

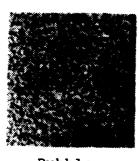
- Autoregressive modeling of the water, grass and sand textures yielded good results. Some of the other textures with more structure and sharp local variations were not reproduced well.
- 2) Autoregressive reproduction of images created after application of the difference operator, with the exception of the water image, yielded generally poor results. As observed in Chapter III, the application of the difference operator produces a seemingly more stationary result, but the edge structure that remained in most of the images after application of the difference operator was in general not reproducible using a purely AR model.
- 3) Application of the inverse filter to the image signal generated by AR model reproduction of the difference operator results yielded smoothed versions of those results. This is similar to what was observed in Section A of this chapter when images were applied to the inverse filter that were not based on the specific data generated by the difference operator.

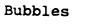
As a final test of the ARIMA model, a 64 × 64 contrast enhanced aerial photograph of trees, with smoother variations and in general less edge structure than the other images tested, was tried. The results are shown in Figure 4-14. Though this image seemed somewhat better adapted to the model, overall the same observations outlined above apply.

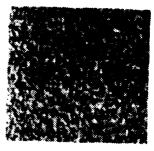
Control of the second of the s











Bark

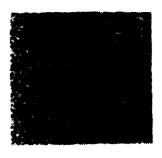


Water



Wall

Figure 4-10 Laplacian Filtered Image Textures Generated Using an AR Model



Sand



Gravel



Grass

Figure 4-11 Laplacian Filtered Image Textures Generated Using an AR Model



Grass



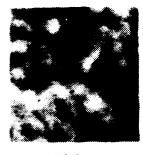
Bark



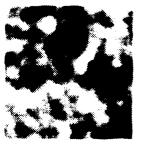
Sand



Water



Bubbles



Wall

Figure 4-12 Image Textures Generated Using an ARIMA Model



Sand

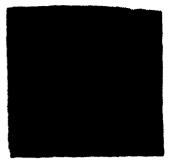


Gravel



Grass

Figure 4-13 Image Textures Generated Using an ARIMA Model



Original Image



Contrast Enhanced



AR Generated Contrast Enhanced Image



Difference Operator Result



AR Generated Difference Operator Result



ARIMA Result

Figure 4-14 Final Test of ARIMA Model on Contrast Enhanced Trees (Magnification X2)

D. SUMMARY

The effectiveness of AR reproduction of image data using white noise input and filter coefficients calculated based on the statistics of the image signal is highly dependent on the nature of the image data. The water image, for example, with its smoothly varying and rather homogeneous nature, was quite well adapted to AR reproduction. Other images with more structure, abrupt variations, and more non-homogeneous characteristics, were not autoregressively reproducible to any great extent.

Using the ARIMA model, it seems that the operation of the inverse filter is very sensitive to the nature of the input data. Input data that are strictly based on the difference operator output can reproduce the original image, as was found in Chapter III. However, the AR model used to generate the inverse filter input (based on the statistics of image signal produced using the difference operator) does not generate image data accurately enough to reproduce images that resemble the real images tested.

V. CONCLUSIONS

This thesis sought to explore experimentally and to understand how linear filtering models could be used to generate texture in images. Of particular interest was the investigation of 2-D ARIMA models to see if they might be of any utility in this effort. Some time was spent exploring separable 2-D models to understand how transfer function pole placement affected image texture characteristics. Image textures generated using these models and applied to the summation filter yielded blurred or smooth textures with seemingly little variety or utility. The ultimate test of model was its ability to reproduce actual The purely AR portion of the model reproduced a few types of actual textures well. However, the full ARIMA model failed to generate image textures that resembled the source images used. Many of the textures had strong edge differences that were not accurately reproducible by the AR model. Also, the summation filter developed seemed very sensitive to deviations in image signal data from that generated by application of the difference operator; that is, the procedure seemed not to be "robust."

Since many of the images tested here have definite edge structure, the difference image had lines which were not reproduced well by the AR model. Correspondingly the

integrated AR model did not reproduce the original image. For this type of image, a combination of a line point process model [Ref. 11] with the integrator, would possibly have been more suitable. The image of trees had not such edge structure and produced somewhat better results. Further experimentation with images of this type and the ARIMA model would perhaps be appropriate.

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APPENDIX A

COMPUTER PROGRAMS, SUBROUTINES, AND FUNCTIONS

Listed below are the names, associated computer systems, and functions of the various computer algorithms used to accomplish this thesis research. All programs were written by the thesis author unless otherwise noted. Program source codes are given at the end of this appendix (except for the MAKFIL* series).

A. PROGRAMS

1. AUTOREG (VAX/VMS FORTRAN)

The program did the following:

- 1) Generated a 128 + 128 zero mean white noise matrix using subroutine PGAUSS.
- 2) Multiplied the white noise by the appropriate image data standard deviation when necessary.
- 3) Converted that matrix into a displayable image file using subroutines SCALE and INTBYTE, when necessary.
- 4) Read filter parameters from an input file into an array.
- 5) Implemented the equation:

$$y(n,m) = -\frac{\prod_{i=1}^{i-1} (-1)}{\prod_{i=0}^{i-1} (n-1,m-1)} \cdot w(n,m) \quad (A.1)$$

using the white noise array and the filter coefficient array as inputs.

- 6) Converted the array result from 4) into a displayable image file using subroutines SCALE and INTBYTE.
- 7) Used subroutine SUBINTFILE to create image data files from filter results for further processing.
- 8) Used subroutine NONC to apply the summation filter to image data when necessary.

2. NONCAUSAL (VAX/VMS FORTRAN)

This program did the following:

- 1) Read filter coefficient values into an array.
- 2) Read image data from an input image file, converted it to integer values using subroutine BYTEINT, calculated the mean from the data, and placed the data into a real array.
- 3) Implemented the equation:

$$y(n,m) = \prod_{i=-L}^{L} \prod_{j=-L}^{I} a_{ij}x(n-i,m-j)$$
 (A.2)

using the image data array and the filter coefficient array.

- 4) Called the subroutine NONC to implement the equation in Step 3) a second time, when necessary.
- 5) Converted the result of Step 3) to a displayable image file using subroutine SCALE and INTBYTE.

3. CONV (VAX/VMX FORTRAN)

This program performs the same basic functions as NONCAUSAL, without having the capability of calling subroutine NONC. It was used for convenience in convolving certain filter structures with certain test images directly.

4. MAKFIL* (VAX/VMS FORTRAN)

This family of programs was used to create various autoregressive and FIR filter coefficient files, using source data manually entered into the program.

5. SPECOR2 (IBM SYSTEM/370 3033 VS FORTRAN 1.4.1)

This program implemented the equations derived in Appendix B and created data files used in developing the corresponding graphs.

6. SPECOR3 (IBM SYSTEM/370 3033 VS FORTRAN 1.4.1)

This program implemented the equations derived in Appendix D and created data files used in developing the corresponding graphs.

7. SPECOR3A (IBM SYSTEM/370 3033 VS FORTRAN 1.4.1)

This program implemented the equations derived in Appendix F and created data files used in developing corresponding graphs.

8. VARIMGS (VAX/VMS FORTRAN)

This program was used to display image data files on the COMTAL (not written by author).

9. PIECE (VAX/VMS FORTRAN)

This program was used to make 128 128 image data files from larger image data files.

10. INTFILE (VAX/VMS FORTRAN)

This program created appropriately formatted integer files from input image data for further processing.

11. TRANS (VAX/VMS FORTRAN)

This program changed the format of filter coefficient data files into a form readable by the image processing programs.

12. NSHP (VAX/VMS FORTRAN)

This program was used to convert quarter-plane autoregressive filter coefficient data to non-symmetric halfplane autoregressive filter coefficient data based on the transformation outlined in Chapter II, Section D.

B. SUBROUTINES

1. PGAUSS (VAX/VMS FORTRAN)

This subroutine, written by C.W. Therrien, was used to generate zero mean, unit variance white noise using RAN (a random number generator function) SQRT, COS, and SIN FORTRAN functions.

2. SCALE (VAX/VMS FORTRAN)

This subroutine takes an image data array and converts it to an integer array with values between an input maximum (MAX) and minimum (MIN) using the following scaling formula:

$$I(i,j) = \frac{(A(i,j)-LOW) \times (MAX-MIN)}{HIGH-LOW} + MIN$$
 (A.3)

A(i,j) is the input image data array, I(i,j) is the output integer array, and HIGH and LOW are the high and low values of A(i,j), respectively (calculated in this subroutine).

This is done to provide appropriate values for image files that will be displayed on the COMTAL Vision One/20, since the gray scale intensity level of each pixel is represented by an 8-bit word. So values possible (in base 10) range from 0 (darkest), to 255 (brightest).

3. INTBYTE AND BYTEINT (VAX/VMS FORTRAN)

These subroutines are necessary since data in an image file are stored in two's complement form. The related variable type in FORTRAN for these values is BYTE. To process image data using FORTRAN implementation of the appropriate formulas, these byte values must be converted to integer (and eventually real using the FLOAT function) form. Results of image processing formulas in real form must be converted to integer (using the INT function) and then byte form to be placed in image data files. INTBYTE converts integer type variables to byte type variables using the following criterion (I is an integer and B is a byte):

If $I \le 127$ and $I \ge 0$ then B = I

If I > 127 and I \leq 255 then B = I-256

BYTEINT converts byte type variables to integer type variables using the following criterion:

If $B \ge -128$ and B < 0 then I = B+256

If $B \ge 0$ and $B \le 127$ then I = B

4. SUBINTFILE (VAX/VMS FORTRAN)

This subroutine performed the same function as INTFILE, but could be called by a program to operate on

processed image data arrays, rather than just image file data inputs.

5. NONC (VAX/VMX FORTRAN)

This subroutine performs essentially the same functions as NONCAUSAL, except that it can be called by a program to operate on an image data array.

C. APL FUNCTIONS

The APL systems on the IBM System/370 3033 and VAX/UNIX were used for matrix manipulations and operations, for graphing filter structures and convolution results, and for calculating autoregressive filter coefficients from image data. All APL functions except MAKMAT were written by C.W. Therrien.

1. MAKMAT (IBM)

This function was used to create the large coefficient matrices (\widetilde{A}) used in calculating the FIR filter coefficients as outlined in Chapter III.

2. CC2 (IBM)

This function was used to circularly convolve the Laplacian FIR filter and its various inverses. Appropriate zero-padding of these filters makes the resulting circular convolution equivalent to linear convolution [Ref. 2:pp. 70-72], which was the desired operation.

3. GETDATA (VAX/UNIX)

This function is used to transfer image data files

4. PUTDATA (VAX/UNIX)

This function is used to transfer filter coefficient data files from an APL workspace to a UNIX subdirectory.

5. MEAN (VAX/UNIX)

This function is used to calculate the mean of an image data file for use in the APL function COVF.

6. COVF (VAX/UNIX)

This function is used to calculate terms in the 2-D covariance function for use in the APL function CORR.

7. CORR (VAX/UNIX)

This function is used to estimate the 2-D covariance function of the image data.

8. MVLEV (VAX/UNIX)

This function is used in APL function FF2DLEV to calculate necessary parameters for the 2-D Levinson recursion from the covariance function of the image data.

9. FF2DLEV (VAX/UNIX)

This function performs the 2-D Levinson recursion to solve for the filter coefficient vector.

AUTOREG

እንደለያቸው መሆነ የትርት መጀመር እና የተመለከተው የተመለከት የተመለከት

```
THIS PROGRAM GENERATES AN IMAGE TEXTURE USING WHITE MOISE AS AN INPUT
      TO AN AUTOREGRESSIVE FILTER AMOSE PARAPMETERS ARE OBTAINED FROM THE
     FILE FILTOEF. SURROUTINE PRAISS IS USED TO GENERATE THE INPUT MHITE
c
     NOISE AND SURROUTINES SCALE AND INTRYTE ARE USED TO PREPARE IMAGE
c
     DATA ARRAYS FOR DISPLAY. SUBPOUTINE SUBINIFILE IS USED IF AN
     INTEGER FILE RESULT IS DESIRED, AND SURROUTINE NONC IS USED IF A
c
     LAPLACIAN INVERSE FILTERING STEP IS NEEDED.
   DEFINE VARIABLES
      byte a(0:127),bim(0:127,0:127)
      integer n.seed.rsize.csize.cml.aml.i.j.iol.row.col.integ(0:127,0:1
      real*8 vall.val2.un(0:127,0:127), run, in(0:127,0:127), coef(0:9,0:9
      (751:0,751:0) tuoni, nim, xem, nev, (1:
c OPEN FILES
      open(unit=1,name="(rathmann.data),undat",type='new',access='direct
      [',recordsize=32,maxrec=129)
      open(unit=2,name="[rathmann.im]final9a.gat",tvoe='new',access='di
     irect', recordsize=32, maxre==129)
      open(3, file='(rathmann.mata) fcdbup.dat', status='old')
C DEFINE PARAMETERS
      seed=1234567
      rsize=127
      csize=127
      0 = 1 = 3
      3-1-3
      0b0.955=xem
      min=27.0d0
C CREATE MHITE NOISE ARRAY
      as 10 is0, rsize
        do 20 j=0.csize=1
          j = 1 = j + 1
          call pgauss(seed, vall, val2)
          wn(i/i)=vall
          wn(i,jol)=val2
       Continue
   10 continue
   SCALE ARRAY AN AND CONVERT TO BYTE FORM
c
       call scale(whinted, was, min)
       call intoyte(integ,bim)
c
   ARTTE THE AMITE VOISE IMAGE ARRAY TO A FILE
c
       do 30 i=0, rsize
         do 40 j=0, ssize
c
           3(i)=5im(i,j)
c
         sontinue
         #rite(1'i+1) (a(n),n=0,rsize)
c
    30 continue
c
   READ FILTER PARA ETERS INTO AN APRAY
      35 50 is0,0ml
        do 50 j=0.0+1
          read(3,55) coef(i,j)
  55
          (51.05r) tearct
  50
        Continue
  50 CONTINUE MULTE NOISE BY REAL IMAGE STANDARD DEVIATION
      var=sort(coef(0,0))
      do 70 isú, reize
        do 90 j=0,csize
          wn(i,j)=wn(i,j)*var
       Continue
```

6.66.666

• 2555555 PX 255555 •

AUTOREG (CONT.)

```
70 continue
   APPLY WHITE HOISE TO THE AUTOREGRESSIVE FILTER
      do 110 n=0, csize
        as 120 ==0,:size
          050-=0.000
          as 130 (=0,sml
             05 140 j=0,071
               if((1.ea.0).and.(j.eq.0)) as to 100
               row=n=i
               col==-i
               if((row.lt.0).or.(col.lt.0)) as to 140
               dsum=(coef(i,j)*in(row,col))+dsum
  140
             continue
  130
          continue
           in(n,m)=(-1.0d0+dsum)+wn(n,m)
        continue
  120
  110 continue
 FILTER THE IMAGE DATA ARRAY USING THE LAPLACTAN INVERSE FILTER
      call nonc(invimout)
   SCALE THE RESULTING IMAGE ARRAY AND CONVERT TO BYTE FORM
      call scale(imput/integ/max/min)
      call intovte(interinim)
   WRITE THE SEVERATED IMAGE INTO A FILE
      do 150 i=0, rsize
        do 160 j=0, ssize
          a(i)=pim(i,i)
        continue write(2'i+1) (a(n),n=0,rsize)
  160
150 continue
c CLOSE FILES
      close(unit=1)
      close(unit=2)
      close(3)
      Stop
      end
```

NONCAUSAL

```
THIS PROGRAM GENERATES AN IMAGE TEXTURE USING A NONCAUSAL FIR FILTER
     AMOSE PARAMETERS ARE UNIATUED FROM 4 DATA FILE. THE FILTER IS
     EPPLIED TO AN IMAGE. SUBROUTINES SCALE, INTEVIE, AND BYTEINT ARE USED TO PREPARE IMAGE DATA ARRAYS FOR DISPLAY. SUBROUTINE NOW.
     IS USED IF AT INTERMEDIATE FILTERING STEP IS DESIRED.
   DEFINE VARIABLES
      byte a(0:127),bim(0:127,0:127)
      integer n,rsize,csize,i,j,row,col,integ(0:127,0:127),fsize,mfsize,
     Sindekleinder2
      real#9 msum.in(0:127,0:127).coef(-10:10,-10:10).msum.mean.dumb(0:1
     127,0:127), high, low, try(-1:128,-1:128), max, mir
 OPEN FILES
      open(unit=1, name=" {rathmann.iml gark.dat', type="old', access="dir
     (851=senakrise=32, maxrec=128)
       ppen(unit=2, name='(rathmann.im)arimbA.dat', tvoe='new', access='dire
c
      int',reportsize=32.maxrec=128)
      open(3, files'(eathwann. dataIncfilco3a.dat', status='old')
```

NONCAUSAL (CONT.)

```
C DEFINE PARAMETERS
      index1=-1
      index2=124
      rsize=127
      csize=127
       fsize=1
       mfsize=-lefsize
      max=255.000
      min=0.030
C READ FILTER PAPAMETERS INTO AN ARRAY
      do 10 immfsize, fsize
        do 20 j=nfsize,fsize
          read(3,25) coef(i,i)
   25
           (S1.05t) tearch
   20
       Continue
   10 continue
  READ IMAGE TO BE FILTERED INTO AN ARRAY AND CONVERT TO INTEGER
      do 30 i=0, rsize
        read(['i+1) (a(n),n=0,127)
        do 40 j=0,csize
        bin(i,j)=a(j)
   40
       Continue
   30 continue
      Call ovteint(pin.intea)
C CONVERT THE INTEGER ARRAY INTO A REAL ARRAY AND COMPUTE THE MEAN
      msum=0.0d0
      high=-10000.0d0
      15==10000.030
      no 30 i=0, rsize
        do 50 j=0.csize
          im(i,i)=float(intex(i,j))
          ¬sun=iπ(i,j)+¬suπ
          if(im(i,j).st.hish) hish=im(i,j)
          if(:r(:,j).)t.|;w: |o.=:m::,j)
       Continue
   50
   50 continue
      (C.851*C.F51)/muzm#re#
      write(4,55)mean,hish,low
   55 format(' priminal image data- mean=',d12.5,' high=',d12.5,' low='d
  SURTRACT THE MEAN FROM THE DATA
      35 70 i=0,127
35 30 j=0,127
          ir(i,j)=in(i,j)=mean
   9.0
       continue
   70 continue
C FILTER THE IMAGE ARRAY
      Szebni, Izebnian 09 cb
        do 100 maindexl/index?
          Otc.C=nuet
          do 110 i=nfsize,fsize
            do 120 janfsize, fsize
              rowin-i
              col==-j
               151 et cc ((0.;1.]co).or.(col.]t.0)) as to 120
               if((raw.ar.127).or.(col.ar.127)) as to 120
              dsum=(in(row.col)*coef(i,i))*dsum
  120
            continue
  110
          continue
          try(n,n)=jsum
       concinue
   90 continue
 CALL SUBROUTINE TO INVERSE FILTER FILTERED IMAGE ARRAY
       call nonc(trv.im)
```

This follows a Caractaria (a)

NONCAUSAL (CONT.)

```
C ADD THE HEAR OF THE TUPUT IMAGE TO THE FILTERED RESULT
      do 145 i=0, esize
        do 155 j=0,:size
          1 m(1 + j) = try(1 + j) + mean
  155
        CONFINUE
  145 continue
C CREATE AV IMAGE DATA FILE
      call subintfile(im)
   SCALE THE RESULTING IMAGE ARRAY AND CONVERT TO BYTE FORM
      call scale(in,integ,max,min)
      call intbyte(interiora)
C ARTTE THE GENERATED IMAGE INTO A FILE
      do 150 i=0.rsize
        do 160 j=0.csize
         a(j)=51=(1,j)
       Continue
        write(2'i+1) (a(n),n=0,rsize)
  150 continue
c CLOSE FILES
     close(unit=1)
      close(unit=2)
     close(3)
      $100
      end
```

CONV

```
THIS PROGRAM GENERATES AN IMAGE TEXTURE USING A MONCAUSAL FIR FILTER MADE PROGRAM OF THE FILTER IS
c
c
      APPLIED TO AT IMAGE. SUBROUTINES SCALE, INTERTE, AND BYTEINT ARE
      JSED TO PREPARE IMAGE DATA ARRIYS FOR DISPLAY.
  DEFINE VARIABLES
       Dyte a(3:127), 0:1(0:127, 0:127)
      integer n.rsize.csize.i.j.row.col.integ(6:127,0:127).fsize.mfsize
real*6 discontine(751:0,751:0).fsize.mfsize
      :27.0:127)
c CPEN FILES
     Doen(unit=1,name='[rathmann.im]arim53.dat',type='old',access='dire
tot',recordsize=32,maxrec=128)
      open(unit=2, name='(rathmann.im)arimo9.jat', type='new',acc+ss='dire
      ict*,recordsize=32,maxrec=124)
      open(3,file='lrathmann.datalconv21.dat',status='old')
C DEFINE PARAMETERS
      rs1ze=127
      cs1ze=127
      fsize=11
       rt512===1+t512e
   YEAR HE CIVI SESTSMARA FELLI CASE
      00 10 immfsize, fsize
        do 20 jantsize, fsize
          read(3,25) spef(1,j)
   25
          for *at ( 121.12)
   20
       Continue
   10 continue
  PERSITE OF TRAVECT CAR AREA TO THE CRESTILL BE LI REAVI CORE
      30 120, rsige
        read(1'++1) (a(n),n=0,127)
        do 40 J=0,05128
        bim(i,j)=a(j)
  40
      Continue
  30 continue
      call syreint(similntes)
```

CONV (CONT.)

\$\$\darks\dar

```
C COUVERT THE INTESER ARRAY INTO A REAL ARRAY AND COMPUTE THE MEAN
       #su==).3d0
       05 50 1=0. rsize
         75 50 1=0,csize
           in(i/i)=float(inteq(i,i))
           #5U#=1#(1,)+#5U#
       CONCINUE
    50 Cantinue
       Tean=nsum/float((rsize+1)+(csize+1))
       write(*,55)mean
    55 format(120.12)
C SURTRACT THE MEAN PHOM THE THAGE ARRAY
       35 70 istoreste
         do an j=U.csize
          ...(1,j)=1n(1,j)=mean
    AU CONTINUE
70 continue
c FILIER THE THAGE ARRAY
       35 30 neO, raige
         do 100 ==0,:size
           asu+=0.∂d0
          ab 110 immfsize,fsize
            do 120 j=nfsize,fsize
              rowanei
               col==-j
               ost es ((0.11.1ca).or.(0.11.wen)) if
               if((row.at.127).or.(col.at.127)) so to 120
              mustr((in(row,sol)+coef(i,i))+dsum
  120
            continue
  110
          continue
          dumo(n.m)=dsum
  100
        continue
   30 continue
  SCALE THE RESULTING IMAGE ARRAY AND CONVERT TO BYTE FORM
      call scale(numo, inteq)
      Call intovte(integrain)
  ARITE THE BEYERATED THAGE INTO A FILE do 150 (=0.nsize
        do 150 j=0,csize
          a(j)=sim(1,j)
  150
        continue
        write(2'i+1) (a(n),n=0,rsize)
 150 continue
c CLOSE FILES
      close(unit=1)
      close(unit=2)
      close(3)
      Stoo
      end
```

SPECOR2

```
C THIS PROGRAM SCLVES EQUATIONS FOR SPECTRAL CONTENT AND CORRELATION
     IN ONE DIRECTION OF A GIVEN AUTOREGRESSIVE IMAGE MODEL. IT MRITES
     THESE RESULTS TO DEVICES 3 AND 4 RESPECTIVELY.
C DEFINE VARIABLES
      INTEGER 1.K
      REAL+8 PI.ALPHA.THETA.H.ALPHAS.SXH.RXK(-49:49).Z.ALPHAK
C DEFINE PARAMETERS
      PI=3.141592654
      ALPHA=0.399999
      ALPHAS=ALPHA==2
C START X AXIS LOOP AND DEFINE X VALUES
      DO 10 1=0.99
       W=(-1.0=PI)+(2.0=PI=(FLOAT(I)/99.0))
C DO SPECTRAL ANALYSIS
       SXH=1.0/(1.0-(2.0=ALPHA=COS(H))+ALPHAS)
       MRITE(5.15)M.SXM
   15 FORMAT(F10.5.1X.F10.5)
   10 CONTINUE
C DC CORRELATION FUNCTION SOLUTION
     DO 23 K+0.49
       ALPHAK #ALPHA##K
       RXK(K) = (ALPHAK/(1.0-ALPHAS))
       I=-1*K
       RXK(I)=RXK(K)
  20 CONTINUE
     DO 30 K=-49.49
       Z#FLOAT(K)
       WRITE(4.25)2.RXK(K)
  25 FORMAT(F10.0.1X.F10.15)
  30 CONTINUE
    STOP
     CHG
```

SPECOR3

```
C THIS PROGRAM SOLVES EQUATIONS FOR SPECTRAL CONTENT AND CORRELATION
    IN ONE DIRECTION OF A GIVEN AUTOREGRESSIVE IMAGE MODEL. IT MRITES
     THESE RESULTS TO DEVICES 3 AND 4 RESPECTIVELY.
C DEFINE VARIABLES
      INTEGER I.J.M
      REAL #8 PI.ALPHA.THETA.ALPHAS.ALSPA.ALPHAS.ALPHA4.THOTH.COSZT.M.A.B
     #,COS2H.SXH1,SXH.K.ALPHAK.KTHETA.ALPHZK.KP2THP.KTHP.INTERM.RXK(-49:
     -49).Z
C DEFINE PARAMETERS
      PI=5.141592654
      ALPHA=0.9
      THETA=0.0=(PI/12.0)
      IF(THETA.EQ.0.0) THETA=0.000001
      ALPHAS . ALPHA . . . S
      AL SPANAL PHAS - AL PHA
      ALPHAS+ALPHA++2
      ALPHAG . ALPHA . . .
      COSST + COS (2.0+THETA)
      THOTH=2.0=THETA
C START X AXIS LOOP AND DEFINE X VALUES
С
      DO 10 1-0.99
        H=(-1.0=PI)-(2.0=PI=(FLOAT(I)/99.0))
C
C DO SPECTRAL ANALYSIS
        A-THETA-H
С
С
        B-THETA-W
С
        COS2H+COS(2.0+H)
c
        SXW1=ALPHAS-(COS(A)=ALSPA)+(ALPHAS+COS2T)-(COS(B)=ALSPA)+(ALPHAS
¢
     ##COSZH)
C
       SXH=1.0/(1.0-ALPHA4+(2.0=SXH1))
C
        WRITE(3.15)W.SXW
C 15 FORMAT(F10.5.1X.F10.5)
C 10 CONTINUE
C DO CORRELATION FUNCTION SOLUTION
        DO 30 J=0.49
          K=FLOAT(J)
          ALPHAK MALPHA MAJ
          KTHETASKSTHETA
          KP2THP=((2.0+K)=THETA)-PI
          KTMP=KTHETA-PI
          INTERM=(COS(KP2TMP)-(ALPHAS=COS(KTMP)))/(1.0+ALPHA4-(2.0=ALPMA
     #S#COS(THOTH)))
          RXK(J)=(ALPHAK/(2.0=(SIH(THETA)==2)))=((COS(KTHETA)/(1.0-ALPHA
     #$1)+INTERM3
C 35
          RXK(J)=(ALPHAK=(2.0+K))/(2.0+((1.0-ALPHAS)==2))
   36
          Le [ --M
          RXK(M)=RXK(J)
   SO CONTINUE
     00 40 3=-49.49
       Z=FLCAT(J)
        HR:TE(4.45)Z.RXK(J)
   45 FORMAT(F10.1.1X.F30.15)
   40 CONTINUE
     STOP
      END
```

SPECOR3A

```
C. THIS PROGRAM SOLVES EQUATIONS FOR SPECTRAL CONTENT AND CORRELATION
     IN ONE DIRECTION OF A GIVEN AUTOREGRESSIVE IMAGE HODEL. IT MRITES
     THESE RESULTS TO DEVICES 3 AND 4 RESPECTIVELY.
C DEFINE VARIABLES
      INTEGER I.J.M
      REAL . FI.A.B.AB.AMBS.AS.BS.AKI.AKZ.BKI.BKZ.INTERM.RXK(-49:49),Z.A
     MK.KPZ.KP1.OPAS.OHAS
C DEFINE PARAMETERS
      PI=3.141592654
      A--0.9
      B--0.89999
      A8-A-B
      AMBS+(A-8)##2
      AS+4**2
      BS+B++2
C START X AXIS LOOP AND DEFINE X VALUES
      DO 10 I=0.99
        W#(-1.0*PI)+(2.0*PI*(FLOAT(I)/99.0))
C DO SPECTRAL ANALYSIS
        COS2M+COS(2.0+M)
        SXH1=1.0-(2.0=(A+8+(AS=8)+(A+8S))=COS(H))+(2.0=A=8+COS2H)
        SXH+1.0/(SXH1+AS+(2.0#A#B)+95+(AS#BS))
        HRITE(3.15)H.SKH
   15 FORMAT(F10.5.1X.F10.5)
   19 CONTINUE
C DO CORRELATION FUNCTION SOLUTION
       DO 20 J=0.49
          IF(A.E0.8) GO TO 35
          AK1=A++(J+1)
          AKC+A++(J+2)
          BK1=B==(J+1)
          8K2×8=+(J+2)
          INTERM=(AKZ/().0-AS)-(((AK)=8)+A=8K)))/(1.0-AB))+(8KZ/(1.0-BS)
         RXK(J)=(1.0/AMBS)=INTERM
         GO TO 36
  35
         AK=A=J
         KP2+FLOAT(J+2)
         KP1=FLOAT(J+1)
         OPAS=1.0-AS
         OMAS=1.0-AS
         RXK(J)=((KPI=KP1=AK)/(2.0=OMAS))-((FUCAT(J)=KP1=AK)/(2.0=OPAS)
    • 1
         RXK(M)+RXK(J)
  30 CONTINUE
     DO 40 Je-49.49
       Z=FLOAT(J)
       WRITE(4.45)Z.RXX(J)
  45 FORMAT(F10.1.1X.F30.15)
  40 CONTINUE
     310°
     CH3
```

PIECE

```
THIS PROGRAM TAKES A TRAVIDA PIECE DUT OF A SIRKSIR IMAGE DATA FILE
      A 10 PLACES THAT DATA (MIT) AM TMAGE FILE
c
c
   DEFINE VANIABLES
      ovte = (0:511), oim(0:5(1,0:5(1),ao(0:127)
   OPEY FILES
c
      open(unit=1,name='(rathmann.im)trees.dat',type='pid',access='pire
      1ct', records) ze=32, wasre:=128}
      ntb'=zeessan'wan'=savi,'iat. bdessailat.nrannjan':=amen.S=jinu)reco
  test ", responsizes to makes cand)
READ IMAGE TO HE FILTERED INTO AN ARRAY
      35 30 1=0,127
         read(1':+1) (a(n),n=0,127)
        as 40 j=0,127
        bim(i,j)=a(j)
        Continue
   40
   30 continue
   ARITE THE DESIRED THASE POPTION INTO A FILE
      35 151 1=53,120
        05 167 1=53,126
          K=1-03
          ap(<)=pim(1,))
  150
        Continue
        write(2'1-62) (ap(n), n= 7.63)
  150 continue
  CLOSE FILES
      close(unit=1)
      close(unit=2)
      $100
      en7
```

INTFILE

```
THIS PROGRAM CHENTES A 128412H OR BAXBO THIESER FILE FROM : 128412H
c
    INPUT IMAGE FILE
c
c
c
¢
  DEFINE VARIABLES
      ovte a(3:127), 0:m(0:127,0:127)
      integer i, j, n, k, l, integ (0:127, 0:127), sum, nigh, low
      real+2 mean
c OPEN FILES
      open{unit=1,name="frathmann.inldliar2.dat",type="gld",access="d
     |irect',recordsize=32,maxr*c=128)
      ppen(2, file="(rarhmann.datal dinmat.dat", status="new")
  YEAR IA UTEL CARALIT BE CT SEEM GASE
      30 i=0,127
        read(1's+11 (a(n),n=0,127)
        do 40 j=0.127
       D1=(i,j)=a(j)
  40
      continue
 30 CONTINUE COVERT THE BATE ARRAY INTO AN INTEGER ARRAY
     call syteint(simintes)
```

Marked Control of the Control of the

INTFILE (CONT.)

```
JAVIMPST OF STIRM ONE ATAC SCAPE SHT SC ESCLAVE MOL OVA HOLF STERMON
      high=
      15==255
      25 31 1=9,127
        do 32 j=0,127
          if(intea(i,j).lt.lo.) low=intea(i,j)
           if(intea(i,j).ar.high) high=integ(i,j)
   32 continue
   31 continue
      weite(4,53)high,low
   33 tarmat(215)
   COMPUTE MEAN OF INTEGER ARRAY AND SUBTRACT IT FROM THE DATA
c
        10 35 i=0,127
         30 45 1=4.127
c
c
           Sun=sun+inted(i,j)
    45
        continue
    35 continue
      nean=float(sun)/(129.0+129.0)
       arite(+,54)sum, mean
    3u tormat(i10,e10.3)
       30 35 i=1,127
         15 45 j=0,127
           inter(i,j)=inter(i,j)=int(mean+0.5)
         continue
   35 continue
ARITE IVIEGES ARRAY INTO 4 128x128 04T4 FILE
       v=124
       1=124
c
       arite(2,391k,1
c
    39 format (2:5)
       30 50 i=0.127
        do 61 (=7,112,15
           #rite(2,70)integ(1,j).integ(1,i+1),integ(i,j+2).integ(i,j+3).i
     -#nteq(1.;+4],:nteq(1,:+5],:nteq(1,;+5),:nteq(1,;+7),:nteq(1,;+7),:nteq(1,:+6),:n
      *rea(1,j+9),integ(i,j+10),integ(1,j+11),integ(1,j+12),integ(1,j+13)
      *,integ(i,i+12),integ(i,j+15)
          format(1515)
   60 continue
   5h continue
CLOSE FILES
      close(unit=1)
      close(unit=2)
      $100
      270
```

TRANS

```
THIS PROGRAM READS DATA DUT DE A FILE IN FREE FORMAT AND CONVERTS

IT ID A PROGRAM READABLE FORMAT.
   integer i,;
   real** a(15)
   open(1/file='[rathmann.dataladwat.dat',status='old']
   open(2/file='(rathmann.dataladwat.dat',status='old')
   read(1/*)i,j,a(1),a(2),a(3),a(4),a(5),a(6),a(7),a(8),a(9),a(10),a(41),a(12),a(12),a(13),a(14),a(15),a(16)
   do 10 k=1,16
        write(2,20)a(k)
20   format(d20,12)
10 continue
   stop
end
```

1000 • KK50066• K099994• 1000024

NSHP

```
THIS PRUSHEM GENERATES AN EMAGE TEXT IRE USING WHITE NOISE AS AN INPUT
c
     TO AV AUTOREGRESSIVE FILTER WHISE PARABETERS ARE INTATUED FROM THE
     FILE FILTOFF, SUARDITIVE PRAUSS IS USED TO SEVERATE THE INPUT WHITE
¢
     NOISE AND SURROUTINES SCALE AND INTERTE ARE ISED TO PREPARE IMAGE
c
     DATA ARRAYS FOR DISPLAY.
c
   DEFINE VARIABLES
      Dyte a(3:127), pim(0:127,0:127)
      integer niseegirsizeicsizeiomlinmlililiplirowicolinteg(0:127,0:1
     127). mom1
      real#3 val1,val2,un(0:127,0:127),rsum,in(0:127,0:127),coef(-2:9,-2
     ::7)
C OPEN FILES
      open(unit=1,name='frathmann.datal.mn.dat',type='nea',a_cess='direct
     ;',records:ze=32,maxrec=128)
      open(unit=2,name='[rathmann.pata]rotate.dat',type='new',access='di
     !rect ', records ize=32, maxre:=128)
      open(3,file='[rathmann.datalfc3a.dat',status='old')
      open(4, file='[rathmann.data]ts.dat', status='nea')
   DEFINE PARAMETERS
      seed=1234567
      rsize=127
      cs1ze=127
      0+1=2
      211=4
      170# | - | TTCF
C CRENTE MAITE NOISE ARRAY
      do 10 i=0, rsize
        oo 20 j=0,csize-1
          jol=j+1
          call prauss(seep,vall,val2)
          wn(i,j)=vall
          ==((:::=1)=val2
   50
       Continue
   10 continue
   SCALE APRAY MY AVO CONVERT TO MYTE FORM
      call scale(wn, inten)
      call intovte(intesisin)
e ARTTE THE WHITE MUISE IMAGE ARRAY TO A FILE
      as 30 i=0.rsize
        do 40 j=0.csize
          a(j)=sim(1.j)
   40
        continue
        write(1'i+1) (a(n),n=0,rsize)
 30 continue - READ FILIER PARAVETERS INTO AN ARRAY
      do 50 immonl.om!
        35 50 j=0,3ml
          read(3,55) coef(1,1)
  55
          format (m20.12)
  hú continue
  50 continue
  APPLY ANTIE NUISE TO THE AUTOREGRESSING FILTER
      10 113 m=0,csize
        as 120 n=0, rsize
          dsu#=0.0d0
          35 130 i=mom1,5m1
            75 140 (=0.5%)
              if((i.es.1).and.(j.eq.1)) as to 140
              ro====i
              c > 1 = = - j
              11((row.1t.1).or.(coi.1t.1)) 20 to 140
              dsun=(coef(i,i)+in(row,col))+isum
 140
            continue
 130
          continue
          + + ( -, + ) = ( -1 + 0 d 0 + d su + ) + wh (n + + )
 120
      continue
 110 continue
```

NSHP (CONT.)

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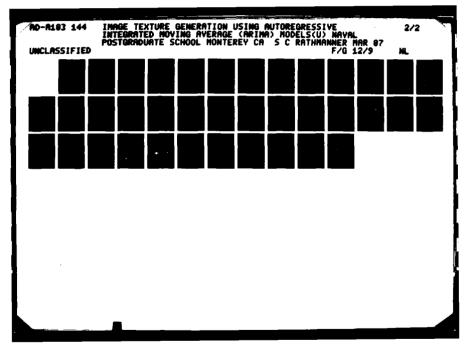
```
C SCALE THE RESULTING IMAGE ARRAY AND CONVERT TO BYTE FORM
      call scale(in,inteq)
      call intovte(integions)
  ARITE THE SEVERATED IMAGE INTO A FILE
      do 150 (=0. maize
        do 160 j=0,csize
         a(j)=pin(i,j)
  150
        Continue
        write(2'i+1) (a(n),n=0,rsize)
  150 continue
c CLOSE FILES
      close(unit=1)
      close(unit=2)
      close(3)
     close(4)
      Stoo
      end
```

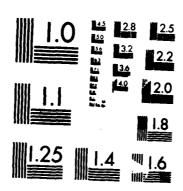
PGAUSS

```
SURROUTINE PSAUSS(K,Z1,Z2)
REAL+R A,B,Z1,Z2
INTESER+4 K
A = STRT(+2.0D0 + ALOG(RAN(K)))
3 = 5.28319500 + RAN(K)
Z1 = A + COS(B)
Z2 = A + SIN(B)
REJERN
ENO
```

SCALE

```
Suprouting scale(arr,inreg, max, min)
  THIS SURROUTIVE ROSLES AN ARRAY TO INTEGER VALUES HETNEEN A SIVEN
     PUPILING CPA NUNIXAN
   DEFINE VARIABLES
      integer integ(0:127,0:127),i,j
      real+3 arr(0:127,0:127),himn,low,mean,max,min
  PERFORA SCALING
      nian=-10000.0m0
      154=17070.010
      do 10 i=0,127
        as 20 j=0,127
          if(arr(i,j).le.lo.) lowzarr(i,j)
          if(arr(i,j).ge.nigh; high=arr(i,j)
          summarr(i,j) tsum
      continue
   10 continue
      nean=su+/(128.0+129.0)
      antre(4,25)niah,low,mean
   25 format(' prescaled image darke high=', d12.5, ' 'h==', d ... ' ---
     (6.5ic.)
C CONTINUE SCALING AND CONVERT TO INTEGER FORM
      do 30 1:0,127
        25 40 j=0,127
          inteq(i,j)=int(f(ann(j,j)=in=j+f(=g,=+in) = = = = =
          integ(1, i)=inteu(1, i)+intimini
      continue
   30 continue
```





MICROCOPY RESOLUTION TEST CHART NATIONAL BUREAU OF STANDARDS-1963-A

(

SCALE (CONT.)

```
C CALCILATE HIGH, LOW, AND MERY OF SCALED IMAGE
      0.0.0ap
      nian=-11000.010
      13-=13070.010
      an 50 i=0.127
        do 50 j=0,127
if(float(inred(i,j)).st.high) high=float(inred(i,j))
          if(floar(inreg(i,i)).lt.low) low=float(integ(i,j))
          sum=float(inter(i,j))+sum
  60 continue
  50 continue
     #gan=sum/(129.0+129.0)
      mrite(*,55)nign,low,mean
  65 format(' scaled image data high=',312.5,' low=',312.5,' mean=',d1
    :2.5)
      return
      end
```

INTBYTE

```
suprouting intovte(intervain)
& THIS SUBTOLINE TAKES AN INTEGER ARRAY AND CONVERTS IT INTO A
    SYTE AIRSY
  DEFINE VARIABLES
     integer i.j.integ(0:127.0:127).n
pyte pin(0:127,0:127)
c PERFORM CONVERSION
     0=0
     do 10 i=0,127
       do 20 j=0.127
         if((integ(i,j).lt.0).pr.(integ(i,j).gt.255)) n=n+1
         #51-=(i,i).it.6) pin(i,i)=-124
         if(intea(i,j).at.?55) bim(i,j)=127
         !integ(i,i)
         if((inteq(i,j).pt.127).and.(inteq(i,j).le.255)) bim(i,j)=
    |integ(1,i)=255
   20 continue
  10 continue
     mrite(*,30) n
   30 format() THE NUMBER OF POINTS OUT OF RANGE IS 1, 15)
     return
     ***
```

BYTEINT

```
suproutine syteint (cim/intes)
  THIS PROBRAM TAKES A RATE ARRAY AND COMPRETS IT INTO AN INTEGER ARRAY
   DEFINE VARIABLES
•
       integer intes(0:127,0:127),i,j,n
  pyte Din(0:127,0:127)
PERFORM CONVERSION
       \alpha = 0
       35 10 i=0,127
         do 20 j=0.127
           if((bi*(i,j).lt.=123).or.(bi*(i,j).st.127)) n=n+1
           if(pin(i,j).lt.-128) inrea(i,j)=-128
if(pin(i,j).at.127) intea(i,j)=127
           if((pim(i,j).ge.~129).and.(pim(i,j).lt.0)) inten(i,j)=
     ! Dim(i. i)+250
           if((bim(i,j).ae.0).ang.(bim(i,j).le.127)) integ(i,j)=bim(i,j)
   20 continue
   10 continue
   write(*,30)n
30 format(* THE NU 48ER OF POINTS OUT OF RANGE IS*,IS)
      return
      end
```

SUBINTFILE

```
Suproutine subintfile(in)
   THIS SUBROUTINE CREATER A 1244124 INTEGER DATA FILE FROM FILTERED
c
     TMAGE DATA
c
c
   DEFINE VARIABLES
       integer i, i, n, x, l, integ(0:127, 0:127), sum
       real** mean.im(0:127,0:127),nigh.low
   DPEN FILES
c
      open(i.file='(nathmann.data)dinnan.dat',sratus='new')
   COMPUTE TISH AND LOW VALUES OF THE IMAGE DATA AND ARITE TO TERMINAL
      hipn=-1.035
      104=1.045
      do 31 1=9,127
        do 32 j=0,127
           if(im(i,j).lt.low) low=imfi,j)
           if(im(i,j).at.high) high=im(i,j)
   32
       Continue
   31 continue
      write(4,33)high,low
   33 format (2020.12)
   COMPUTE MEAN OF INTEGER ARRAY AND SUBTRACT IT FROM THE DATA
c
c
       90923
c
       35 i=1,127
         10 45 1=3,127
c
           Sun=sun+integ(i,j)
c
        continue
    35 continue
c
      mean=float(sum)/(128.0+128.0)
c
c
       write(+,34)sum, mean
    34 format(i10,e10.3)
       35 i=0.127
c
         30 45 j=0,127
c
           integ(i,j)=integ(i,j)=int(mean+0.5)
c
        sportinue
    35 continue
```

SUBINTFILE

```
INTFILE

do)

Dola File

((*,iol).inten(i,;iol).inten(i,iol)...

((*,iol).inten(i,;iol).inten(i,;ol)...

((*,iol).inten(i,;iol).inten(i,;iol)...

((*,iol).inten(i,;iol)...

((*,i
         CONVERT IMAGE DATA INTO INTESER FORM
                   35 45 i=0,127
                        do 55 i=0,127
                             integ(i,j)=int(in(i,j)+0.5d0)
                        Continue
           45 continue
         MRITE INTEGER ARRAY THTO A 124x128 DATA FILE
                  x=128
                  1=128
                  write(1,39)4,1
          39 format (215)
                  as 50 i=0.127
                       do 50 j=0,112,14
                            weite(1,77)intea(i,j),intea(i,j+1),intea(i,j+2),intea(i,j+3),i
               *ntea(i,j+4),intea(i,i+5),intea(i,j+5),intea(i,j+7),intea(i,j+5),in
               *teq(i,j+9),inteq(i,j+10),inteq(i,j+11),inteq(i,j+12),inteq(i,j+13)
               *rintes(iri+14),intes(iri+15)
         70
                          format(1015)
         60
                     Continue
         50 continue
 c CLOSE FILES
                Close(unit=1)
                 return
                223
                                                                                               NONC
                Suproutine nonc(try,in)
       THIS SUBJUITIVE TAKES AN IMAGE APRLY AND FILTERS IT WITH A
c
             WONCAUSAL FIR FILTER.
c
c
       DEFINE VARIABLES
                integer norsize.csize.i,i,row.cnl,fsize,mfsize.indexl,index2
                real+3 gsum, nm(0:127,0:127), coef(-10:10,-10:10), msum, mean, gump(-1:
             :128,-1:129),nign,low,try(0:127,0:127)
      OPEN FILES
               open(1, file='(rathmann.data)ncfilco21.dat', status='old')
       DEFINE PARAMETERS
                index1=0
                index2=127
               rsize=127
               cs12e=127
               fsize=19
               misize=-letsize
       READ FILTER PARAMETERS INTO AH ARRAY
               Jo 10 ismisize.fsize
                    do 20 jamtsizertsize
                         read(1,25) coef(i,i)
       25
                          ($1,05b) servet
       20
                    continue
       10 continue
      TAKE THE LYPUT ARRAY AND COMPUTE THE MEAN, MIGH, AND LOW VALUES
```

55 format(' inout image array= mean=',dl2.5,' high=',dl2.5,' low=',dl

900.C=quer high=-19000.030 15-=13030.040

continue 50 continue

60

.2.51

do 50 isindest, inter2 Suerni.luetni=i 0¢ co msum=try(i,j)+msum

write(+,55) mean, high.lo.

if(trv(i,j).gt.high) high=try(i,j) (i,i)vry=wcl (wol.11.(i,i)vry)li

mean=ms_m/float((innex2-index1+1)+(innex2-index1+1))

NONC (CONT.)

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```
C FILTER THE THAGE ARRAY
      Sascni, Lastnian 00 cb
        do 100 mainmest, index?
          010.0=eueb
          do 110 immfsize, fsize
            do 120 i=nfsize,fsize
              rowen-i
              col===;
               if((rom.lr.index1).or.(col.)t.index1)) so to 120
               of ((Sagreinter2).or.(Sagreinter2)) as to 120
              rust+((i,f)) aco+((co,wcn) yrs) = rusb
  120
            continue
  110
          continue
          in(7, n)=1sun
  100
        continue
   90 continue
   CONVERT THE 130x130 ARRAY INTO A 128x128 ARRAY, IF NECESSARY
c
       do 130 i=0,127
c
         do 140 j=0,127
¢
           (i,i) oruc=(i,i) rr
c
   140
        continue
c
   130 continue
      close(1)
      return
      end
```

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MAKMAT

```
P R+makmat.OIO:A:C:I
OIO----
A- (1 11 66 PU
AF0:0:0:0]+4
AF0:0:1]+-4
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START1+1
LOOF: I+I+1
START1+START1+(I-1)
START1+START1+(I+1)
POM4---
                                                                                                                                                                                         ♥ B+mAKMAT:DIG:A:C:D:E:F;G,H,I,J:K.L;H,O
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GETDATA

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PUTDATA

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CORR

```
# RIFSIZE CORP F-DID:R-U;K,L,K

# FUNCTION TO ESTIMATE ID COVAR AMOD FUNCTION FOR THACT -

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MVLEV

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A ARTHULEV RPT.RT.N.NB:NDIM.AA1.AF:AB:AF1.AR1:EF.ER.GAMF.GAMB:DELT,I.DIG

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| FUNCTION TO BENERATE TERMS IN 20 COVARIANCE FOR FOR IMAGE

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APPENDIX B

DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE TWO POLE AUTOREGRESSIVE MODEL

Power Spectrum

$$|H(e^{j\omega})|^2 = H(e^{j\omega}) \cdot (H(e^{j\omega}))^* = H(e^{j\omega})H(e^{-j\omega})$$

In this case:

$$H(e^{j\omega}) = \frac{1}{1+\alpha e^{-j\omega}}$$
 $H(e^{-j\omega}) = \frac{1}{1+\alpha e^{j\omega}}$

Calculating $H(e^{j\omega})H(e^{-j\omega})$:

$$\frac{1}{(1+\alpha e^{-j\omega})(1+\alpha e^{j\omega})} = \frac{1}{1+\alpha e^{-j\omega} + \alpha e^{j\omega} + \alpha^2} = \frac{1}{1+2\alpha\cos(\omega) + \alpha^2}$$

The final result is:

$$S_{\mathbf{y}}(\omega) = |H(e^{j\omega})|^2 = \frac{1}{(1+\alpha^2)+2\alpha\cos(\omega)}$$
 (B.1)

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Autocorrelation Function

Starting with H(z) for this case:

$$H(z) = \frac{1}{1 + \alpha z^{-1}}$$

Per Ref. 7:p. 158:

$$Z^{-1}[H(z)] = h(n) = (-\alpha)^n \cdot u(n)$$
 for $\alpha < 1$ ($u(n)$ is the unit step function)

Per Ref. 5:pp. 391-395, for the white noise input case:

$$R_{\mathbf{Y}}(\ell) = \sum_{n=-\infty}^{\infty} h(n) \cdot h(n-\ell)$$
 (B.2)

Substituting h(n) above into Eq. B.2

$$R_{\mathbf{y}}(\ell) = \sum_{n=\ell}^{\infty} (-\alpha)^{n} \cdot (-\alpha)^{n-\ell} \qquad \ell \geq 0$$

$$= \frac{1}{(-\alpha)^{\ell}} \sum_{n=\ell}^{\infty} (-\alpha)^{2n} \qquad \ell \geq 0$$

The summation term may also be expressed as:

$$\sum_{n=\ell}^{\infty} (-\alpha)^{2n} = \sum_{n=0}^{\infty} (-\alpha)^{2n} - \sum_{n=0}^{\ell-1} (-\alpha)^{2n}$$

Per Ref. 12:p. 8, the summation terms on the right are equal to:

$$\sum_{n=0}^{\infty} (-\alpha)^{2n} = \frac{1}{1 - (-\alpha)^2} (\alpha < 1) \qquad \sum_{n=0}^{\ell-1} (-\alpha)^{2n} = \frac{1 - (-\alpha)^{2\ell}}{1 - (-\alpha)^2} (\alpha < 1)$$

As a result:

$$\sum_{n=\ell}^{\infty} (-\alpha)^{2n} = \frac{1}{1 - (-\alpha^2)} - \frac{1 - (-\alpha)^{2\ell}}{1 - (-\alpha)^2} = \frac{(-\alpha)^{2\ell}}{1 - (-\alpha)^2} \quad (\alpha < 1)$$

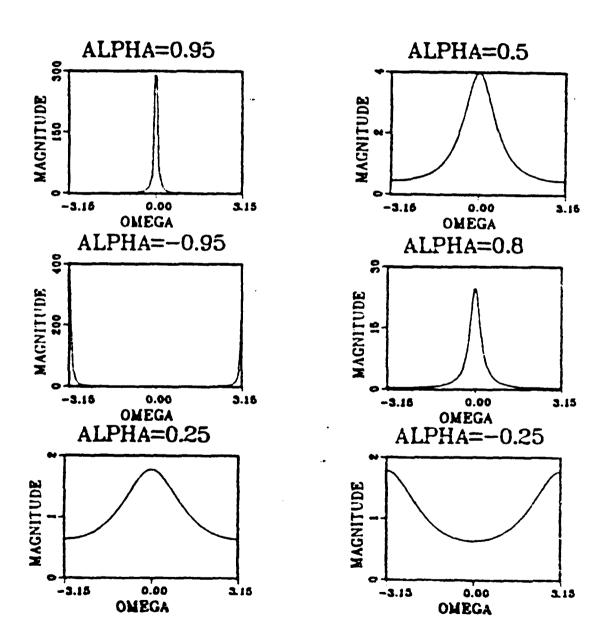
Substituting and using $(-\alpha)^2 = \alpha^2$ yields:

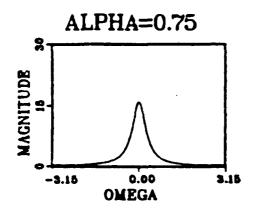
$$R_{\mathbf{Y}}(\ell) = \frac{1}{(-\alpha)^{\ell}} \cdot \frac{(-\alpha)^{2\ell}}{1 - \alpha^2} = \frac{(-\alpha)^{\ell}}{1 - \alpha^2} \quad \ell \ge 0 \text{ and } \alpha < 1$$
 (B.3)

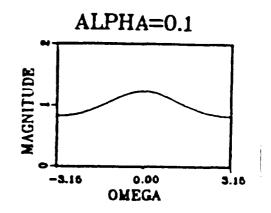
APPENDIX C

GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE TWO POLE AUTOREGRESSIVE MODEL

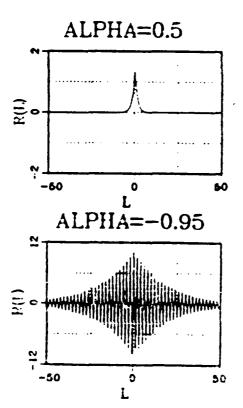
Power Spectrum

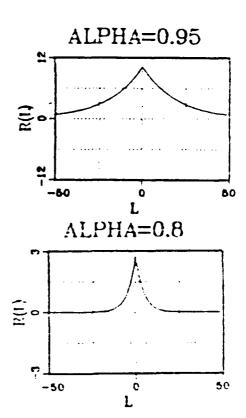


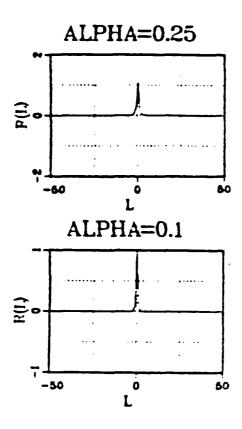


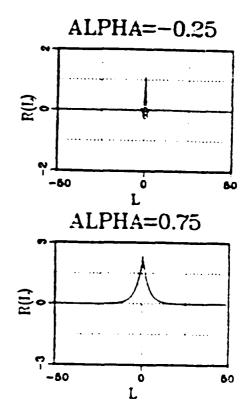


Autocorrelation Function









APPENDIX D

DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL

Power Spectrum

$$|H(e^{j\omega})|^2 = H(e^{j\omega}) \cdot H(e^{-j\omega})$$

In this case:

$$\begin{split} \mathbf{H}(\mathbf{e}^{-\mathbf{j}\omega}) &= \frac{1}{(1-\alpha\mathbf{e}^{-\mathbf{j}\theta}\mathbf{e}^{\mathbf{j}\omega})(1-\alpha\mathbf{e}^{\mathbf{j}\theta}\mathbf{e}^{\mathbf{j}\omega})} &= \frac{1}{1-\alpha\mathbf{e}^{\mathbf{j}(\theta+\omega)}-\alpha\mathbf{e}^{-\mathbf{j}(\theta-\omega)}+\alpha^{2}\mathbf{e}^{\mathbf{j}2\omega}} \\ \mathbf{H}(\mathbf{e}^{\mathbf{j}\omega}) &= \frac{1}{(1-\alpha\mathbf{e}^{-\mathbf{j}\theta}\mathbf{e}^{-\mathbf{j}\omega})(1-\alpha\mathbf{e}^{\mathbf{j}\theta}\mathbf{e}^{-\mathbf{j}\omega})} &= \frac{1}{1-\alpha\mathbf{e}^{-\mathbf{j}(\theta+\omega)}-\alpha\mathbf{e}^{\mathbf{j}(\theta-\omega)}+\alpha^{2}\mathbf{e}^{-\mathbf{j}2\omega}} \end{split}$$

Multiplying the above expressions yields:

$$\begin{split} \text{H}(\mathbf{e}^{\mathbf{j}\omega})\text{H}(\mathbf{e}^{-\mathbf{j}\omega}) &= \frac{1}{\mathbf{1} - \alpha \mathbf{e}^{\mathbf{j}(\theta - \omega)} - \alpha \mathbf{e}^{-\mathbf{j}(\theta + \omega)} + \alpha^2 \mathbf{e}^{-\mathbf{j}2\omega} - \alpha \mathbf{e}^{\mathbf{j}(\theta + \omega)} + \alpha^2 \mathbf{e}^{\mathbf{j}2\theta} + \alpha^2} \\ &- \alpha^3 \mathbf{e}^{\mathbf{j}(\theta - \omega)} - \alpha \mathbf{e}^{-\mathbf{j}(\theta - \omega)} + \alpha^2 \mathbf{e}^{-\mathbf{j}2\theta} - \alpha^3 \mathbf{e}^{-\mathbf{j}(\theta - \omega)} \\ &+ \alpha^2 \mathbf{e}^{\mathbf{j}2\omega} - \alpha^3 \mathbf{e}^{\mathbf{j}(\theta + \omega)} - \alpha^3 \mathbf{e}^{-\mathbf{j}(\theta + \omega)} + \alpha^4 \end{split}$$

Combining terms:

$$\begin{split} \text{H}(e^{j\omega}) \cdot \text{H}(e^{-j\omega}) &= \frac{1}{1 + 2\alpha^2 + \alpha^4 - \alpha^3 (e^{-j(\theta - \omega)} + e^{j(\theta - \omega)}) + \alpha^2 (e^{j2\theta} + e^{-j2\theta})} \\ &- \alpha^3 (e^{j(\theta + \omega)} + e^{-j(\theta + \omega)}) + \alpha^2 (e^{j2\omega} + e^{-j2\omega}) \\ &- \alpha (e^{j(\theta - \omega)} + e^{-j(\theta - \omega)}) - \alpha (e^{j(\theta + \omega)} + e^{-j(\theta + \omega)}) \end{split}$$

Using Euler's relation and combining terms:

$$H(e^{j\omega}) \cdot H(e^{-j\omega}) = \frac{1}{1+\alpha^4+2(\alpha^2-\cos(\theta-\omega)[\alpha^3+\alpha]+\alpha^2\cos(2\theta)-\cos(\theta+\omega))}$$

$$\times [\alpha^3+\alpha]+\alpha^2\cos(2\omega))$$

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Using $\cos(\theta - \omega) + \cos(\theta + \omega) = 2\cos(\theta)\cos(\omega)$, and since $S_{\mathbf{V}}(\omega) = H(e^{j\omega}) \cdot H(e^{-j\omega}) \cdot \sigma^2$, with $\sigma^2 = 1$ the final result is:

$$S_{\mathbf{Y}}(\omega) = \frac{1}{1+\alpha^4+2(\alpha^2-2[\alpha^3-\alpha]\cos(\theta)\cos(\omega)+\alpha^2(\cos(2\theta)+\cos(2\omega)))}$$
 (D.1)

Autocorrelation Function

$$H(z) = \frac{1}{1 - \alpha e^{j\theta} z^{-1} - \alpha e^{-j\theta} z^{-1} + \alpha^2 z^{-2}} = \frac{1}{1 - (e^{j\theta} + e^{-j\theta}) z^{-1} + \alpha^2 z^{-2}}$$

Using Euler's relation:

$$H(z) = \frac{1}{1-2\alpha \cos(\theta) z^{-1}+\alpha^2 z^{-2}} |z| > |\alpha|$$

Per Ref. 7:pp. 204-216, partial fraction expansion can be used to find the inverse Z transform. To do so H(z) can be expressed in the form:

$$H(z) = \frac{1}{(1-\alpha e^{-j\theta}z^{-1})(1-\alpha e^{j\theta}z^{-1})} = \frac{z}{(z-\alpha e^{-j\theta})(z-\alpha e^{j\theta})}$$

Using the partial fraction expansion and table look up [Ref. 7:p. 158] yields:

$$h(n) = \frac{\alpha^n}{\sin(\theta)} \cos(n\theta + \theta - \frac{\pi}{2}) \cdot u(n) \quad \text{for } \alpha < 1$$
 (D.2)

Since $\cos(\theta - \frac{\pi}{2}) = \sin(\theta)$, the final expression for h(n) is:

$$h(n) = \frac{\alpha^n}{\sin(\theta)} \sin((n+1)\theta) \cdot u(n) \qquad \alpha < 1$$
 (D.3)

For simplicity in further derivation of $R_{\mathbf{Y}}(\ell)$ based on h(n), Eq. D.2 will be used.

Using the expression for the autocorrelation function of a random process represented by the above filter with a white noise input [Ref. 4:pp. 391-395]:

$$R_{\mathbf{Y}}(\ell) = \sum_{n=-\infty}^{\infty} h(n) \cdot h(n-\ell) = \frac{\alpha^{-\ell}}{\sin^{2}(\theta)} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot \cos(n\theta + \theta - \frac{\pi}{2})$$

$$\cdot \cos((n-\ell)\theta + \theta - \frac{\pi}{2})$$
(D.4)

 $n=\ell$ in the summation index since h(n) is causal. ℓ is assumed to be greater than zero here. For $\ell<0$, $R_{Y}(\ell)=R_{Y}(-\ell)$ by symmetry of the autocorrelation function [Ref. 5:p. 388], so we can proceed assuming only positive values of ℓ .

Using the trigonometric identity for a product of cosines:

$$R_{\mathbf{Y}}(\ell) = \frac{\alpha^{-\ell}}{\sin^{2}(\theta)} \sum_{n=\ell}^{\infty} \alpha^{2n} \left[\frac{1}{2} \cdot \cos(\ell\theta) + \frac{1}{2} \cos(2n\theta - \ell\theta + 2\theta - \pi) \right] \quad \ell \geq 0$$

$$P_{\mathbf{Y}}(\ell) = \frac{\cos(\ell\theta) \cdot \alpha^{-\ell}}{2\sin^{2}(\theta)} \cdot \sum_{n=\ell}^{\infty} \alpha^{2n} + \frac{\alpha^{-\ell}}{2\sin^{2}(\theta)} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot (2n\theta - \ell\theta + 2\theta - \pi)$$
 (D.5)

ℓ ≥ 0

Using $\sum_{n=\ell}^{\infty} \rightarrow \sum_{n=0}^{\infty} -\sum_{n=0}^{\ell-1}$ and standard geometric progression identities [Ref.]2:p. 8]:

$$\sum_{n=0}^{\infty} \alpha^{n} = \frac{1}{1-\alpha} \quad \text{and} \quad \sum_{n=0}^{\ell-1} \alpha^{n} = \frac{1-\alpha^{\ell}}{1-\alpha}$$

For the first term in $R_{\mathbf{V}}(\ell)$

$$\sum_{n=\ell}^{\infty} \alpha^{2n} = \sum_{n=0}^{\infty} \alpha^{2n} - \sum_{n=0}^{\ell-1} \alpha^{2n} = \frac{1}{1-\alpha^2} - \frac{1-\alpha^{2\ell}}{1-\alpha^2}$$

$$= \frac{\alpha^{2\ell}}{1-\alpha^2}$$
(D.6)

For the second term in $R_{V}(l)$:

let
$$\phi = - \ell\theta + 2\theta - \pi$$

Using Euler's relation:

$$\cos(2n\theta+\phi) = \frac{e^{j(2n\theta+\phi)} + e^{-j(2n\theta+\phi)}}{2}$$

$$\sum_{n=\ell}^{\infty} \alpha^{2n} \cdot \cos(2n\theta+\phi) = \frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} [e^{j(2n\theta+\phi)} + e^{-j(2n\theta+\phi)}]$$

$$= \frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot e^{j(2n\theta+\phi)} + \frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot e^{-j(2n\theta+\phi)}$$
(D.7)

For large n, it is evident that the α^{2n} term will tend to make the term in each sum approach 0 for $\alpha < 1$, and thus ensures convergence and a closed form expression for each term. Pursuing the mathematics required to find this closed form expression we have:

$$\frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot e^{j(2n\theta+\phi)} = \frac{e^{j\phi}}{2} \sum_{n=\ell}^{\infty} (\alpha e^{j\theta})^{2n} = \frac{e^{j\phi}}{2} [\sum_{n=0}^{\infty} (\alpha e^{j\theta})^{2n} - \sum_{n=0}^{\ell-1} (\alpha e^{j\theta})^{2n}]$$

$$\frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot e^{j(2n\theta + \phi)} = \frac{e^{j\phi}}{2} \left[\frac{1}{1 - (\alpha e^{j\theta})^2} - \frac{1 - (\alpha e^{j\theta})^{2\ell}}{1 - (\alpha e^{j\theta})^2} \right]$$

$$= \frac{\alpha^{2\ell} \cdot e^{j(2\theta\ell + \phi)}}{2(1 - (\alpha e^{j\theta})^2)} \tag{D.8}$$

For the conjugate term we must have:

$$\frac{1}{2} \sum_{n=\ell}^{\infty} \alpha^{2n} \cdot e^{-j(2n\theta + \phi)} = \frac{\alpha^{2\ell} \cdot e^{-j(2\theta\ell + \phi)}}{2(1 - (\alpha e^{-j\theta})^2)}$$
 (D.9)

Next a common denominator must be found to sum these two terms:

$$\begin{split} \frac{\alpha^{2\ell} \cdot \mathrm{e}^{\mathrm{j}\,(2\theta\ell+\varphi)}}{2\,(1-(\alpha\mathrm{e}^{\mathrm{j}\,\theta})^2)} \cdot \frac{(1-(\alpha\mathrm{e}^{-\mathrm{j}\,\theta})^2)}{(1-(\alpha\mathrm{e}^{-\mathrm{j}\,\theta})^2)} \\ &= \frac{\alpha^{2\ell}\mathrm{e}^{\mathrm{j}\,(2\theta\ell+\varphi)} - \alpha^{\,(2\ell+2)}\,\mathrm{e}^{\mathrm{j}\,(2\theta\ell-2\theta+\varphi)}}{2\,(1-(\alpha\mathrm{e}^{\mathrm{j}\,\theta})^2-(\alpha\mathrm{e}^{-\mathrm{j}\,\theta})^2+\alpha^4)} \end{split}$$

$$\begin{split} \frac{\alpha^{2\ell} \cdot \mathrm{e}^{-\mathrm{j}\,(2\theta\ell+\phi)}}{2\,(1-(\alpha\mathrm{e}^{-\mathrm{j}\,\theta})^2)} & \cdot \frac{(1-(\alpha\mathrm{e}^{\mathrm{j}\,\theta})^2)}{(1-(\alpha\mathrm{e}^{\mathrm{j}\,\theta})^2)} \\ & = \frac{\alpha^{2\ell}\mathrm{e}^{-\mathrm{j}\,(2\theta\ell+\phi)} - \alpha^{\,(2\ell+2)}\,\mathrm{e}^{-\mathrm{j}\,(2\theta\ell-2\theta+\phi)}}{2\,(1-(\alpha\mathrm{e}^{\mathrm{j}\,\theta})^2-(\alpha\mathrm{e}^{-\mathrm{j}\,\theta})^2+\alpha^4)} \end{split}$$

By Euler's relation:

$$1-(\alpha e^{j\theta})^2-(\alpha e^{-j\theta})^2+\alpha^4 = 1-2\alpha^2\cos(2\theta)+\alpha^4$$

Adding the terms with the common denominator yields:

$$\frac{\alpha^{2\ell} \left[e^{j(2\theta\ell+\phi)} - \alpha^2 e^{j(2\theta\ell-2\theta+\phi)} + e^{-j(2\theta\ell+\phi)} - \alpha^2 e^{-j(2\theta\ell-2\theta+\phi)} \right]}{2(1-2\alpha^2 \cos(2\theta) + \alpha^4)}$$

Again using Euler's relation the above expression reduces to:

$$\frac{\alpha^{2\ell} \left[\cos\left(2\theta\ell + \phi\right) - \alpha^2 \cos\left(2\theta\ell - 2\theta + \phi\right)\right]}{1 - 2\alpha^2 \cos\left(2\theta\right) + \alpha^4} \tag{D.10}$$

which is the sum of the last two terms in Eq. (D.7). Substituting Eq. (D.10) and Eq. (D.6) into Eq. (D.5) yields:

$$R_{\mathbf{Y}}(l) = \frac{\cos(l\theta) \cdot \alpha^{-l} \left[\alpha^{2l} \left[1 - \alpha^{2l}\right]\right]}{2\sin^{2}(\theta) \left[1 - \alpha^{2l} \left[\cos(2\theta l + \phi) - \alpha^{2} \cos(2\theta l - 2\theta + \phi)\right]\right]}$$

$$+ \frac{\alpha^{-l} \left[\alpha^{2l} \left[\cos(2\theta l + \phi) - \alpha^{2} \cos(2\theta l - 2\theta + \phi)\right]\right]}{1 - 2\alpha^{2} \cos(2\theta) + \alpha^{4}}$$
(D.11)

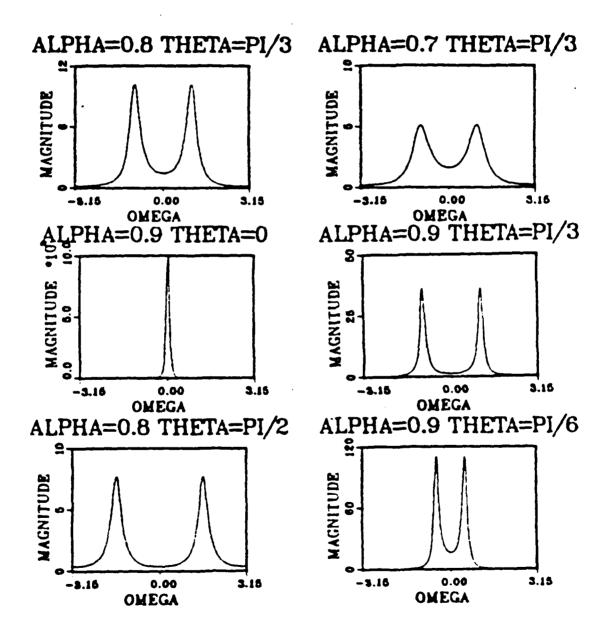
Combining and canceling terms and substituting for ϕ and noticing that the same result must hold for ℓ < 0 we have:

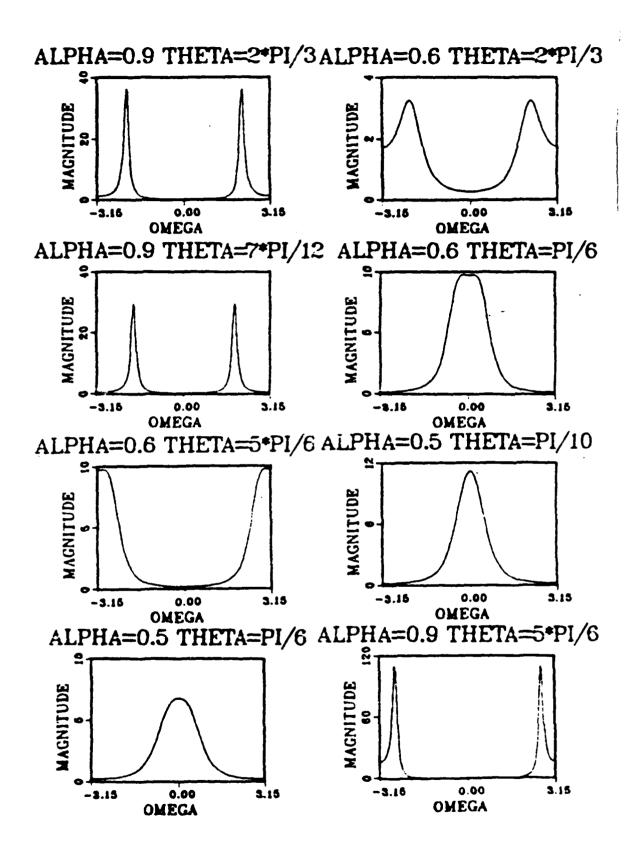
$$R_{\mathbf{y}}(\ell) = \frac{\alpha^{\ell}}{2\sin^{2}(\theta)} \left[\frac{\cos(|\ell|\theta)}{1-\alpha^{2}} + \frac{\cos((2+|\ell|)\theta-\pi) - \alpha^{2}\cos(|\ell|\theta-\pi)}{1-2\alpha^{2}\cos(2\theta) + \alpha^{4}} \right]$$
 (D.12)

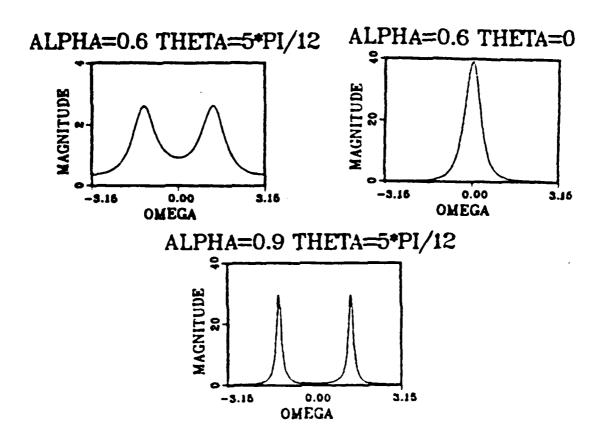
APPENDIX E

GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL

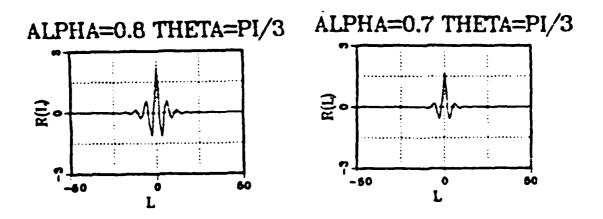
Power Spectrum

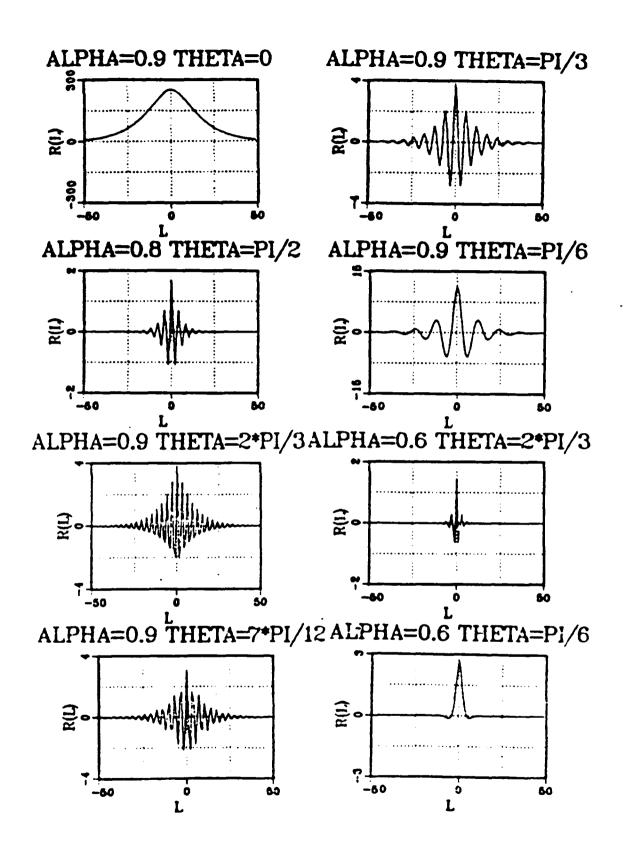


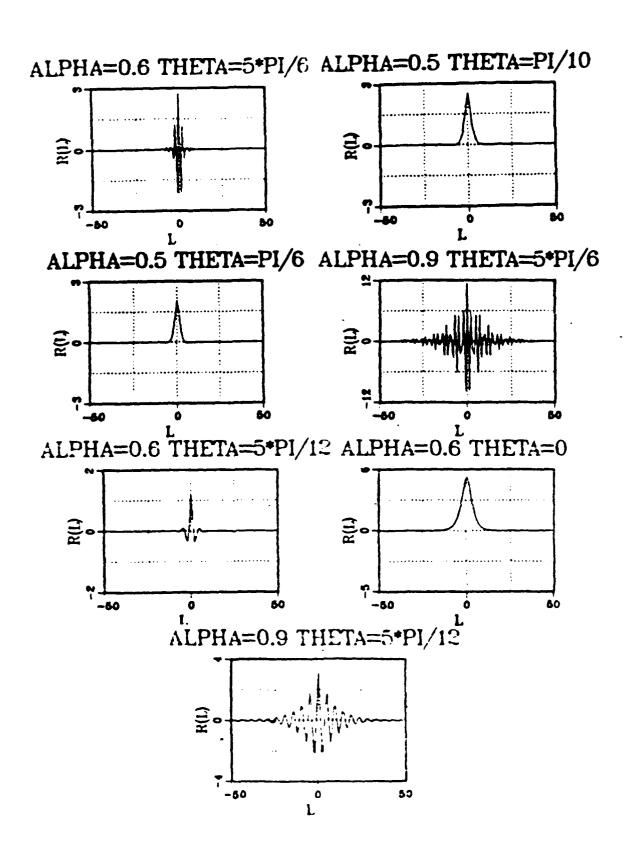




Autocorrelation Function







APPENDIX F

DERIVATION OF THE POWER SPECTRUM AND AUTOCORRELATION FUNCTION FOR THE FOUR POLE AUTOREGRESSIVE MODEL (WITH TWO POLES ON THE REAL AXIS)

Power Spectrum

Using initial results from Appendix D with the necessary modifications (including $\theta = 0$) we have:

$$\begin{split} \mathtt{H}(\mathbf{e}^{-\mathbf{j}\omega}) &= \frac{1}{1-\alpha_{\mathbf{a}}\mathbf{e}^{\mathbf{j}\omega}-\alpha_{\mathbf{b}}\mathbf{e}^{\mathbf{j}\omega}+\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}\mathbf{e}^{\mathbf{j}2\omega}} \\ \mathtt{H}(\mathbf{e}^{\mathbf{j}\omega}) &= \frac{1}{1-\alpha_{\mathbf{a}}\mathbf{e}^{-\mathbf{j}\omega}-\alpha_{\mathbf{b}}\mathbf{e}^{-\mathbf{j}\omega}+\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}\mathbf{e}^{-\mathbf{j}2\omega}} \end{split}$$

Therefore:

$$\begin{split} \mathtt{H}(\mathtt{e}^{\mathtt{j}\omega}) \cdot \mathtt{H}(\mathtt{e}^{-\mathtt{j}\omega}) &= \frac{1}{1 - \alpha_a e^{-\mathtt{j}\omega} - \alpha_b d^{-\mathtt{j}\omega} + \alpha_a \alpha_b e^{-\mathtt{j}2\omega} - \alpha_a e^{\mathtt{j}\omega} + \alpha_a^2 + \alpha_a \alpha_b} \\ &- \alpha_a^2 \alpha_b e^{-\mathtt{j}\omega} - \alpha_b e^{\mathtt{j}\omega} + \alpha_a \alpha_b + \alpha_b^2 - \alpha_a \alpha_b^2 e^{-\mathtt{j}\omega} \\ &+ \alpha_a \alpha_b e^{\mathtt{j}2\omega} - \alpha_a^2 \alpha_b e^{\mathtt{j}\omega} - \alpha_a \alpha_b^2 e^{\mathtt{j}\omega} + \alpha_a^2 \alpha_b^2 \end{split}$$

Combining terms and using Euler's relation:

$$H(e^{j\omega}) \cdot H(e^{-j\omega}) = \frac{1}{1 - (\alpha_a + \alpha_b + \alpha_a^2 \alpha_b + \alpha_a \alpha_b^2) \cdot 2\cos(\omega) + 2\alpha_a \alpha_b \cos(2\omega) + \alpha_a^2} + 2\alpha_a \alpha_b + \alpha_b^2 + \alpha_a^2 \alpha_b^2$$

Assuming $\sigma^2 = 1$ and since $S_y(\omega) = H(e^{j\omega}) \cdot H(e^{-j\omega}) \cdot \sigma^2$, the final result is:

$$S_{\mathbf{Y}}(\omega) = \frac{1}{1-2(\alpha_{\mathbf{a}}+\alpha_{\mathbf{b}}+\alpha_{\mathbf{a}}^{2}\alpha_{\mathbf{b}}+\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}^{2})\cos(\omega)+2\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}\cos(2\omega)+\alpha_{\mathbf{a}}^{2}+2\alpha_{\mathbf{a}}\alpha_{\mathbf{b}}} + \alpha_{\mathbf{b}}^{2}+\alpha_{\mathbf{a}}^{2}\alpha_{\mathbf{b}}^{2}$$

$$+ \alpha_{\mathbf{b}}^{2}+\alpha_{\mathbf{a}}^{2}\alpha_{\mathbf{b}}^{2}$$
(F.1)

Autocorrelation Function

$$H(z) = \frac{1}{1 - \alpha_a e^{j\theta} z^{-1} - \alpha_b e^{-j\theta} z^{-1} + \alpha_a \alpha_b z^{-2}}$$

Letting $\theta = 0$

$$H(z) = \frac{1}{1 - (\alpha_a + \alpha_b) z^{-1} + \alpha_a \alpha_b z^{-2}} = \frac{z^2}{z^2 - (\alpha_a + \alpha_b) z + \alpha_a \alpha_b}$$

$$|z| > |\alpha|$$

Expanding in terms of partial fractions we have:

$$H(z) = \frac{\alpha_a}{\alpha_a - \alpha_b} \cdot \frac{z}{z - \alpha_a} + \frac{\alpha_b}{\alpha_b - \alpha_a} \cdot \frac{z}{z - \alpha_b}$$

This corresponds to the impulse response

$$h(n) = \left(\frac{\alpha_a}{\alpha_a - \alpha_b} \cdot \alpha_a^n + \frac{\alpha_b}{\alpha_b - \alpha_a} \cdot \alpha_b^n\right) \cdot u(n) \qquad \alpha < 1$$
 (F.2)

Proceeding as in Appendix D:

$$R_{\mathbf{y}}(\ell) = \sum_{n=-\infty}^{\infty} h(n) \cdot h(n-\ell) = \sum_{n=\ell}^{\infty} (\frac{\alpha_{\mathbf{a}}^{n+1}}{\alpha_{\mathbf{a}}^{-\alpha_{\mathbf{b}}}} + \frac{\alpha_{\mathbf{b}}^{n+1}}{\alpha_{\mathbf{b}}^{-\alpha_{\mathbf{a}}}}) (\frac{\alpha_{\mathbf{a}}^{n+1-\ell}}{\alpha_{\mathbf{a}}^{-\alpha_{\mathbf{b}}}} + \frac{\alpha_{\mathbf{b}}^{n+1-\ell}}{\alpha_{\mathbf{b}}^{-\alpha_{\mathbf{a}}}})$$

$$(\ell > 0)$$

$$R_{Y}(\ell) = \sum_{n=\ell}^{\infty} \left[\frac{\alpha_{a}^{2n+2-\ell}}{(\alpha_{a}^{-\alpha}\alpha_{b}^{-1})^{2}} + \frac{\alpha_{a}^{n+1-\ell}\alpha_{b}^{n+1-\ell}}{(\alpha_{a}^{-\alpha}\alpha_{b}^{-1})(\alpha_{b}^{-\alpha}\alpha_{a}^{-1})} + \frac{\alpha_{a}^{n+1-\ell}\alpha_{b}^{n+1}}{(\alpha_{b}^{-\alpha}\alpha_{a}^{-1})(\alpha_{a}^{-\alpha}\alpha_{b}^{-1})} + \frac{\alpha_{a}^{2n+2-\ell}\alpha_{b}^{-1}\alpha_{a}^{-1}\alpha_{b}^{-1}}{(\alpha_{b}^{-\alpha}\alpha_{a}^{-1})^{2}} + \frac{\alpha_{a}^{2n+2-\ell}\alpha_{b}^{-1}\alpha_{a}^{-1}\alpha_{b}^{-1}\alpha_{a}^{-1}\alpha_{b}^{-1}\alpha_$$

Since $(\alpha_a - \alpha_b) = -(\alpha_b - \alpha_a)$:

$$(\alpha_a - \alpha_b)^2 = (\alpha_b - \alpha_a)^2 = -(\alpha_a - \alpha_b)(\alpha_b - \alpha_a) = -(\alpha_b - \alpha_a)(\alpha_a - \alpha_b)$$

So:

$$R_{\mathbf{y}}(\ell) = \frac{1}{(\alpha_{\mathbf{a}} - \alpha_{\mathbf{b}})^{2}} [\alpha_{\mathbf{a}}^{2-\ell} \sum_{n=\ell}^{\infty} \alpha_{\mathbf{a}}^{2n} - \alpha_{\mathbf{a}} \alpha_{\mathbf{b}}^{1-\ell} \sum_{n=\ell}^{\infty} (\alpha_{\mathbf{a}} \alpha_{\mathbf{b}})^{n}$$
$$- \alpha_{\mathbf{b}} \alpha_{\mathbf{a}}^{1-\ell} \sum_{n=\ell}^{\infty} (\alpha_{\mathbf{b}} \alpha_{\mathbf{a}})^{n} + \alpha_{\mathbf{b}}^{2-\ell} \sum_{n=\ell}^{\infty} \alpha_{\mathbf{b}}^{2n}] \quad (\ell \ge 0)$$

Continuing with the same principles and assumptions as in Appendix D, we have:

$$\sum_{n=\ell}^{\infty} \alpha^{2n} = \sum_{n=0}^{\infty} \alpha^{2n} - \sum_{n=0}^{\ell-1} \alpha^{2n} = \frac{1}{1-\alpha^2} - \frac{1-\alpha^{2\ell}}{1-\alpha^2} = \frac{\alpha^{2\ell}}{1-\alpha^2}$$

$$\sum_{n=\ell}^{\infty}\alpha^n = \sum_{n=0}^{\infty}\alpha^n - \sum_{n=0}^{\ell-1}\alpha^n = \frac{1}{1-\alpha} - \frac{1-\alpha^{\ell}}{1-\alpha} = \frac{\alpha^{\ell}}{1-\alpha}$$

Making the appropriate substitutions in the expression for $R_{\mathbf{y}}(\ell)$, we have:

$$\begin{split} R_{\mathbf{y}}(\ell) &= \frac{1}{(\alpha_{\mathbf{a}} - \alpha_{\mathbf{b}})^2} [\alpha_{\mathbf{a}}^{2-\ell} (\frac{\alpha_{\mathbf{a}}^{2\ell}}{1 - \alpha_{\mathbf{a}}^2}) - \alpha_{\mathbf{a}} \alpha_{\mathbf{b}}^{1-\ell} (\frac{\alpha_{\mathbf{a}}^{\ell} \alpha_{\mathbf{b}}^{\ell}}{1 - \alpha_{\mathbf{a}} \alpha_{\mathbf{b}}}) \\ &- \alpha_{\mathbf{b}} \alpha_{\mathbf{a}}^{1-\ell} (\frac{\alpha_{\mathbf{a}}^{\ell} \alpha_{\mathbf{b}}^{\ell}}{1 - \alpha_{\mathbf{a}} \alpha_{\mathbf{b}}}) + \alpha_{\mathbf{b}}^{2-\ell} (\frac{\alpha_{\mathbf{b}}^{2\ell}}{1 - \alpha_{\mathbf{b}}^2}) \quad (\ell \geq 0) \end{split} \tag{F.4}$$

Combining terms yields the final expression:

$$R_{Y}(l) = \frac{1}{(\alpha_{a} - \alpha_{b})^{2}} \left[\frac{\alpha_{a}^{2+l}}{1 - \alpha_{a}^{2}} - \frac{\alpha_{a}^{l+1} \alpha_{b} + \alpha_{a}^{l+1} \alpha_{b}^{l+1}}{1 - \alpha_{a}^{\alpha} \alpha_{b}} + \frac{\alpha_{b}^{2+l}}{1 - \alpha_{b}^{2}} \right]$$
 (F.5)

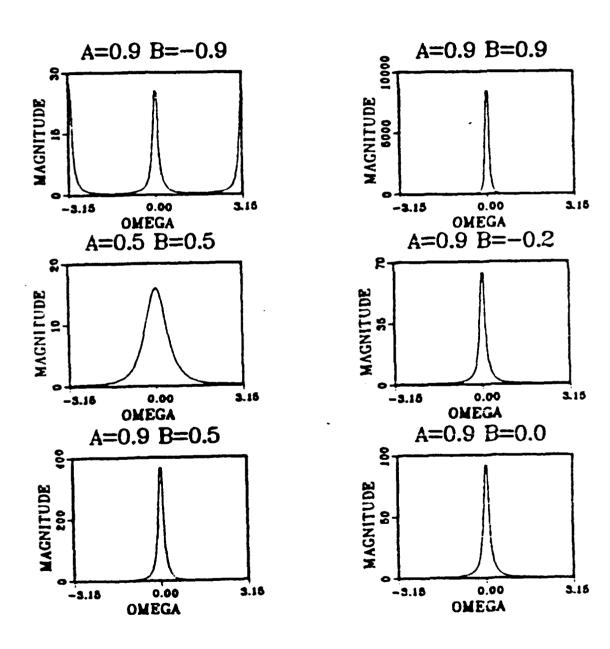
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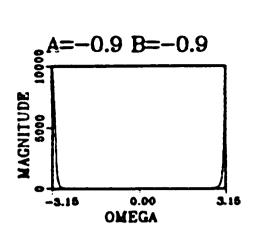
APPENDIX G

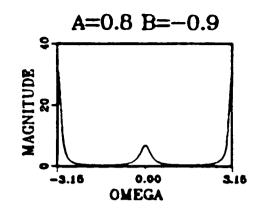
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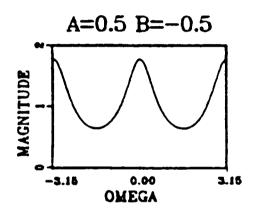
GRAPHICAL RESULTS FOR THE POWER SPECTRUM AND AUTO-CORRELATION FUNCTION FOR THE FOUR POLE AUTO-REGRESSIVE MODEL (WITH TWO POLES ON THE REAL AXIS)

Power Spectrum

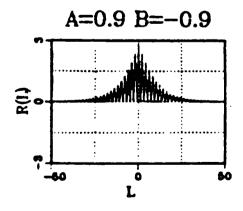


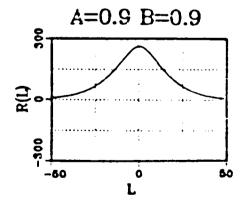


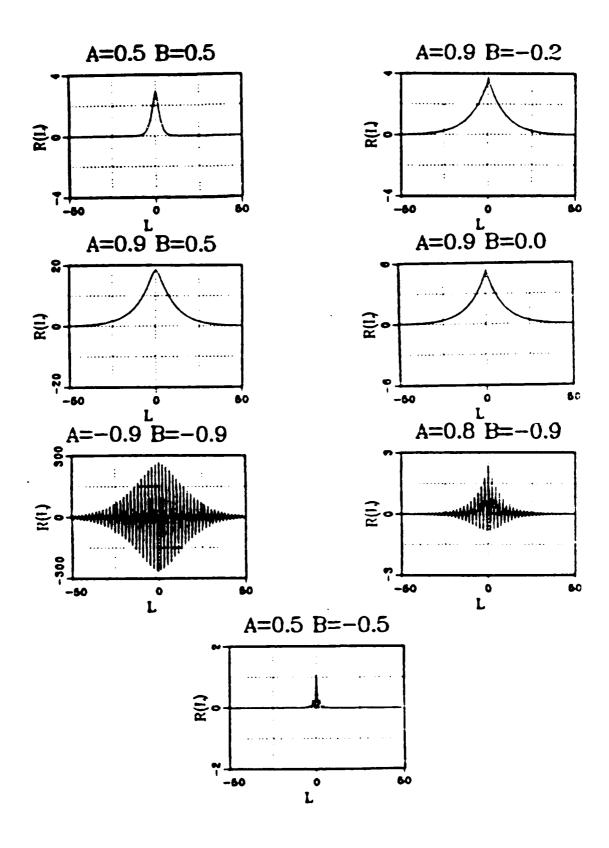




Autocorrelation Function

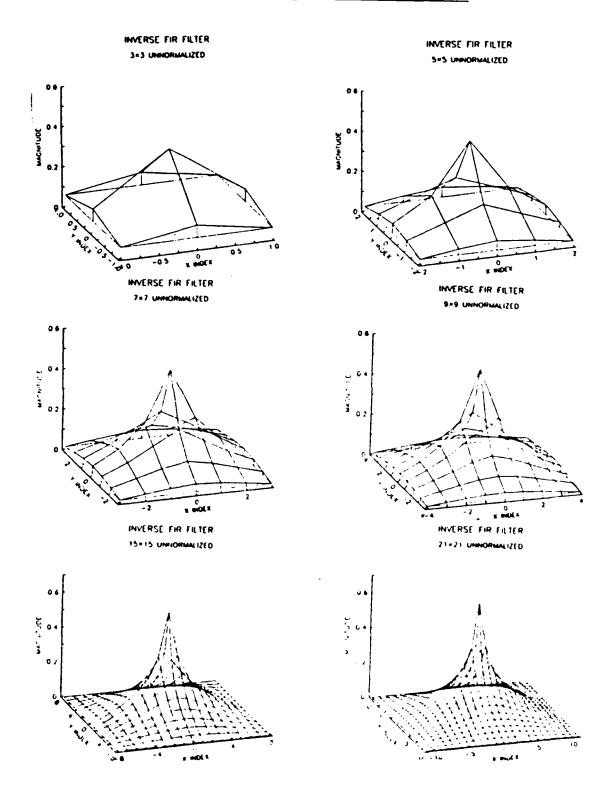




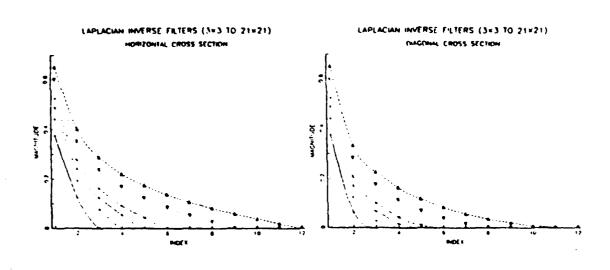


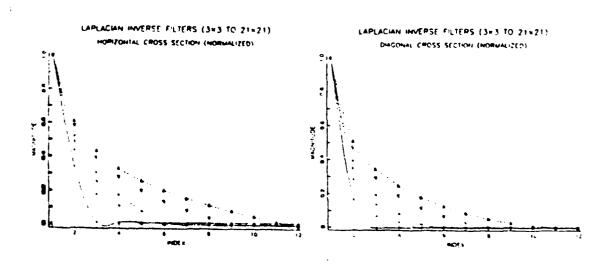
APPENDIX H

LAPLACIAN INVERSE FILTER FORMS



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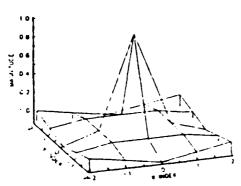


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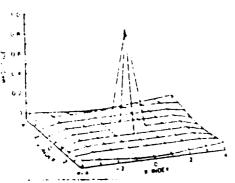
APPENDIX I

CONVOLUTION OF LAPALACIAN DIFFERENCE OPERATOR AND VARIOUS SIZE FIR INVERSE FILTERS

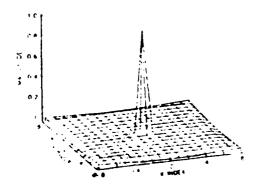
CONVOLUTION OF LAPLACIAN AND ITS 3×3 INVERSE



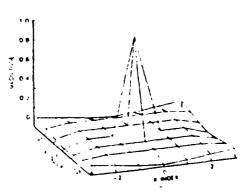
CONVOLUTION OF LAPLACIAN AND ITS 7×7 INVERSE



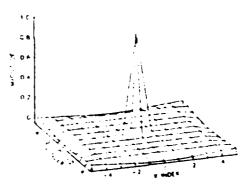
CONVOLUTION OF LAPLACIAN AND ITS 15+15 MYERS!



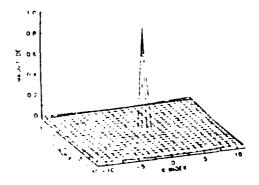
CONVOLUTION OF LAPLACIAN AND ITS 5+5 HIVERSE



CONVOLUTION OF LAPLACIAN AND ITS 949 HIVERSE



CONVOLUTION OF LAPLACIAN AND ITS 21=21 HIVE BET



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